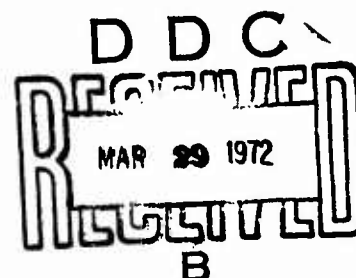


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**NAVAL POSTGRADUATE SCHOOL**  
**Monterey, California**  
**93940**



1 FEB 1972 NPS-6IMd 72021A

Amplitude Modulation Of An  
Acoustic Wave  
Propagating Near The Sea Surface

W.J. Smith, Jr. Ocean Physics Group  
Physics Department

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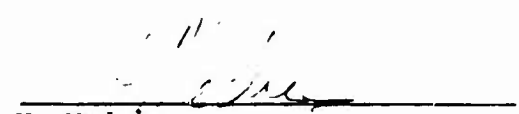
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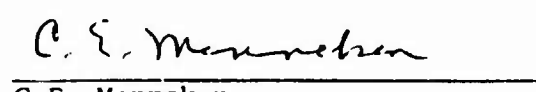
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## I. INTRODUCTION

### A. ACOUSTIC AMPLITUDE MODULATION IN A RANDOM MEDIUM

Consider an underwater acoustic wave propagated at a constant frequency and a fixed amplitude in isotropically homogeneous, undisturbed water with no scatterers. In the far field of the transducer, the wave will arrive at the receiver unchanged except for a diminution due to spherical spreading loss and absorption. The amplitude will not vary about the peak-to-peak value received. For low values of absorption, this type of acoustic wave is adequately described by the wave equation:

$$\nabla^2 p = \frac{1}{c^2} \frac{\partial^2 p}{\partial t^2}$$

Near the ocean surface, a wave will encounter random inhomogeneities which cause a varying speed of sound resulting in refraction. The wave equation must therefore be modified to include temporal and spatial changes in the speed of sound. Such changes may be expressed as a change in the index of refraction  $\mu(\vec{r}, t)$ , where  $u(\vec{r}, t)$  is the ratio of average sound speed to local sound speed:

$$\mu(\vec{r}, t) = \frac{c_0}{c(\vec{r}, t)}$$

which may be expressed as:

$$\mu(\vec{r}, t) = 1 + \alpha n(\vec{r}, t)$$

where  $\alpha$  is the root-mean-square variation of the index of refraction from unity, and  $n(\vec{r}, t)$  is the normalized variation of  $\mu$  from unity. The quantity  $\alpha$  has been shown to be a small number, equal to  $1.18 \times 10^{-4}$  in laboratory experiments using heated water (Stone and Mintzer, 1965) and to  $2.55 \times 10^{-5}$  in ocean experiments. (Whitmarsh, Skudrzyk and Urlick, 1957.)

The effect of the random variation of the index of refraction on an acoustic wave may be shown by a variation in the amplitude of a received signal, noticable as a low frequency modulation of the carrier frequency. Such modulation may be expressed as a fraction of the maximum undisturbed amplitude of the received signal as shown below:

$$\text{instantaneous fractional modulation} = \left[ \frac{V_{MAX} - V_{MOD}}{V_{MAX}} \right]$$

where

$V_{MAX}$  = the unmodulated voltage value

$V_{MOD}$  = the instantaneous modulated voltage value

The inhomogenieties which cause the variations in the refractive index, and ultimately result in amplitude modulation, are temporal and spatial variations in temperature and salinity micro-structure, random bubble populations, particulate matter, varying forms of oceanic life and those inhomogenieties caused by man. Each factor will vary with location and ambient conditions in the ocean. The effect on sound transmission may also be a function of sound frequency.

Particle velocity in the ocean has a great effect on the above parameters, generally in determining their temporal and spatial

distribution. There are two main types of oceanic particle velocity, that due to convective turbulence and that due to the effect of surface waves. Convective turbulence is fairly slow and affects the distribution of temperature and salinity. Such concepts as temperature "patches" and lenticular volumes of near uniform temperature are due to convective turbulence.

Turbulence arising from the effect of surface wave action is generally responsible for random variations in populations of scatterers and for small scale variations in temperature and salinity distribution.

Distributions and variations of all parameters affected by wave turbulence would show a lessening effect with increasing depth due to the  $e^{-kz}$  decay of wave amplitude. Convective turbulence could exist at all depths near the surface and its dependence on depth is unknown. These mechanisms are always present near the surface of the ocean and are a principal factor in the modulation of sound. They also lead to problems in analysis. For example, a stationary temperature patch might show temporal characteristics due to displacement across a sensor with wave action.

Statistical analysis of the amplitude modulation occurring in inhomogeneous transmission has centered about the coefficient of variation, CV, defined by Stone and Mintzer (1961):

$$(CV)^2 = \frac{(\langle P^2 \rangle - \langle P \rangle^2)}{\langle P \rangle^2}$$

where CV = the coefficient of variation

P = the amplitude of the received signal

$\langle \rangle$  = the ensemble average.

The same coefficient of variation is defined equivalently by Campanella and Favret (1969) as:

$$(CV)^2 = \left[ \frac{\langle S(t) - \langle S(t) \rangle \rangle}{\langle S(t) \rangle} \right]^2$$

where  $S(t)$  is the amplitude of the received signal.

The relationship between the coefficient of variation as defined by Stone and Mintzer,  $(CV)$ , and the fractional modulation as defined in this thesis is shown in Appendix I.

The relation between  $(CV)^2$  and the RMS variation of the index of refraction  $\alpha$ , the range  $r$  and the size of the inhomogenieties has been variously described, as by Stone and Mintzer (1965) depending on the (presumed) shape of the spatial correlation function of the index of refraction  $R(\rho)$  where:

$$R(\rho) = \langle n(\vec{r}) n(\vec{r} + \rho) \rangle$$

If  $R(\rho)$  is Gaussian,

$$R(\rho) = \exp(-\rho^2/a^2)$$

then the coefficient of variation may be derived for the conditions that

$$kr \gg 1$$

$$ka \gg 1$$

$$\text{and, } r/ka^2 \ll 1$$

as:

$$(CV)^2 = \frac{1}{2} \sqrt{\pi} k_o^2 \alpha^2 a r$$

where  $k_o$  = space wave number of the sound frequency  
 $\alpha$  = RMS variation of the index of refraction  
 $a$  = radius of the scattering inhomogeneity  
 $r$  = transducer to receiver range.

If  $R(\rho)$  is exponential;

$$R(\rho) = \exp(-\rho/a)$$

then the numerical factor in the expression for  $(CV)^2$  above would be unity. Although an exponential correlation function would not satisfy the requirements of non-zero slope at  $\rho = 0$ , the experimental work (Stone and Mintzer, 1961 and 1965, and Campanella and Favret, 1969) seems to indicate an exponential correlation function. The dilemma can be resolved by using a modified exponential function to satisfy the  $\rho = 0$  condition.

The temporal correlation function for the variation of the index of refraction,

$$M(\tau) = \langle n(t) n(t+\tau) \rangle$$

has been compared to the temporal correlation function for the variation of sound pressure amplitudes,

$$\Phi(\tau) = \langle S(t) S(t+\tau) \rangle$$

and the relation:

$$M(\tau) = \Phi(\tau)$$

which was derived theoretically by Mintzer (1954) and was shown to exist experimentally by Campanella and Favret (1969) in a controlled laboratory environment using convective temperature differences to generate a varying index of refraction.

Experimental work in the ocean environment has included propagation at frequencies from 800 Hz (R.M. Kennedy, 1969) and 60 kHz (Whitmarsh et al., 1957) at distances of 1000 yds (Whitmarsh et al.) and twenty-five miles. (Kennedy.)

Kennedy found the probability distribution function of the logarithm of the amplitude fluctuations to be normally distributed, and an exponential temporal correlation function with an average decorrelation time to  $1/e$  in 12 minutes. There was no attempt to take measurements of the environment during his experiment. Whitmarsh, Skudrzyk and Urick found that the fluctuation in amplitude tended to increase with range, and decried a lack of good information concerning the microthermal structure of the ocean.

#### B. SCOPE AND INTENT OF THE EXPERIMENT

The purpose of the experiment was to measure and analyze the temporal variation in the fractional modulation of an acoustic wave over an oceanographically defined volume in the ocean. The temporal characteristics of the environmental factors of temperature, salinity, turbulent velocity, sound speed and wave height were measured simultaneously with fluctuations in sound amplitude.

A simple one-transducer and hydrophone arrangement over a short transmission length of two meters was adequate to measure variations in the amplitude of the transmitted signal. The short range was desirable because of the need for system rigidity and an adequate definition of the environmental factors occurring during each twenty-minute run. The length of each run was the minimum time necessary to statistically define the low frequency environmental factors which caused the amplitude modulation.

The acoustical and oceanographic analysis at each depth was concerned with the statistical and temporal characteristics. The results desired were the mean and variance, the probability density function, the temporal correlation and the power spectral density of each variable.

The acoustic results will later be cross-correlated with the varying environmental factors to determine the relative effect of each parameter upon the amplitude modulation.

## II. DESIGN OF THE EXPERIMENT

### A. DETERMINATION OF THE CONFIGURATION OF THE EXPERIMENTAL APPARATUS

#### 1. Conceptual Design of the Experiment

In order to measure environmentally caused amplitude modulation over a clearly defined volume in the ocean, a high frequency acoustic carrier wave was directed at a receiving hydrophone mounted upon the acoustic axis at a separation of two meters. The received signal was processed for the amplitude modulation occurring in each twenty-minute run.

In support of the acoustic investigation, an electromagnetic flowmeter, three thermistors, a Bissett-Berman apparatus with a salinometer and a temperature probe and a Ramsey probe with a velocimeter and a wave-height probe were mounted about the acoustic apparatus to define the environment during each run.

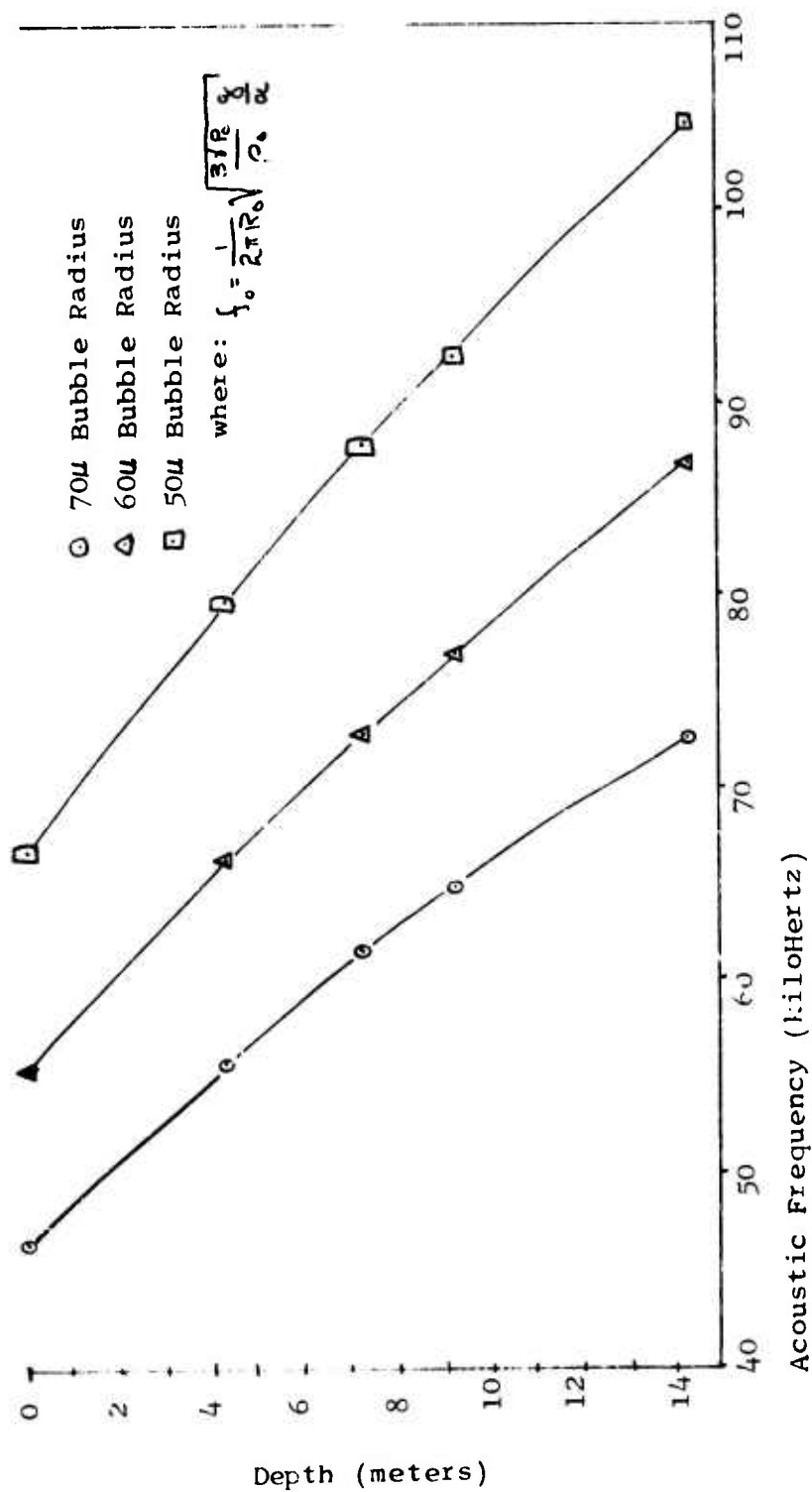
#### 2. The Acoustic Carrier Wave

The carrier wave utilized in the experiment was a continuous-wave 60 kHz constant amplitude signal. This frequency was selected in order to maximize the effect of random bubble populations which were anticipated to be greatest near the ocean surface.

As may be seen from Figure II-1, the 60 kHz carrier frequency was near resonance for the most populous bubble sizes near the surface. The resonance curves were calculated on the basis of the formula:

$$f_0 = \frac{1}{2\pi R_0} \sqrt{\frac{3\gamma P_0}{\rho_0} \frac{g}{a}}$$





BUBBLE RESONANT FREQUENCY VERSUS DEPTH

Figure II-1

where  $f_o$  = resonance frequency (Hz)  
 $R_o$  = radius of the bubble (cm)  
 $P_o$  = ambient pressure at the bubble depth, equals  $10^6$   
 dynes/cm<sup>2</sup> at the surface with  $10^6$  dynes/cm<sup>2</sup> added  
 for each 10 meters of depth  
 $\rho_o$  = density of water (1.0 gram/cm<sup>3</sup>)  
 $\gamma$  = ratio of specific heats (assumed to be 1.4)  
 $g\alpha^{-1}$  = correction for surface tension and thermal  
 conductivity equals 1.0296 for 60 kHz

### 3. Determination of the Acoustic Range

The most stringent requirements for the distance between transducer and hydrophone arose from the desirability that

- (a) the hydrophone be in the far field of the transducer.
- (b) the volume of water under investigation be locally defined in terms of the measurable parameters of temperature, salinity, speed of sound and surface wave height.
- (c) the mounting of the hydrophone be stiff enough to preclude significant changes of sound amplitude through the sound beam.
- (d) the path length be long enough to observe measurable amplitude modulation.

The compromise transducer to hydrophone distance was set at two meters. Checking requirement (a), for the available source (F-27) using the criterion  $r = a^2/\lambda$ , and assuming an average sound speed of 1500 meters a second;

$$f = 60 \times 10^4 \text{ Hz}$$

$$\lambda = c/f = (1500/6) \times 10^{-4} = 2.50 \times 10^{-2} \text{ meters}$$

$$a = 0.127 \text{ meters, where } a \text{ is the radius of the transducer face}$$

then, the far field begins at (approximately)

$$r = a^2/\lambda = 0.01613/0.0250 = 0.645 \text{ meters}$$

Thus the far field begins well within the two meter distance between the transducer and the hydrophone.

For requirements (b) and (d) the two meter distance was selected to be long enough to yield measurable amplitude modulation, yet short enough to allow oceanographic measurements on a definable volume of water. The data from the runs proved this selection to be justified.

The alignment of the receiving hydrophone with the acoustic axis of propagation was an important consideration. A narrowly defined beam pattern reduced reflection from the apparatus, but a radius of error had to be calculated in order to establish the effect of minute movement of the receiver (requirement (c)).

Pressure amplitude as a function of angle was calculated from the formula below, where only the last term is important;

$$P = j\rho_0 c \left[ \frac{ka^2}{2r} \right] u_0 e^{j(\omega t - kr)} \left[ \frac{2J_1(ka \sin \theta)}{ka \sin \theta} \right]$$

For one per cent modulation, an angular displacement of  $0.45^\circ$  would have to occur, two meters, the radius of movement would be;

$$\theta = 0.45^\circ$$

$$\sin \theta = x/2 \text{ meters}$$

$$x = 2(0.0075)$$

$$x = 0.0150 \text{ meters}$$

The radius of movement would have to be 1.50 centimeters, more movement than the mounting would allow. As a final test at

the scene of the experiment, a diver "twanged" the hydrophone supporting wire with no observable result in amplitude modulation. It is estimated that no more than 0.1% modulation could be caused by movement of the hydrophone by ocean forces.

#### 4. Design of the Apparatus

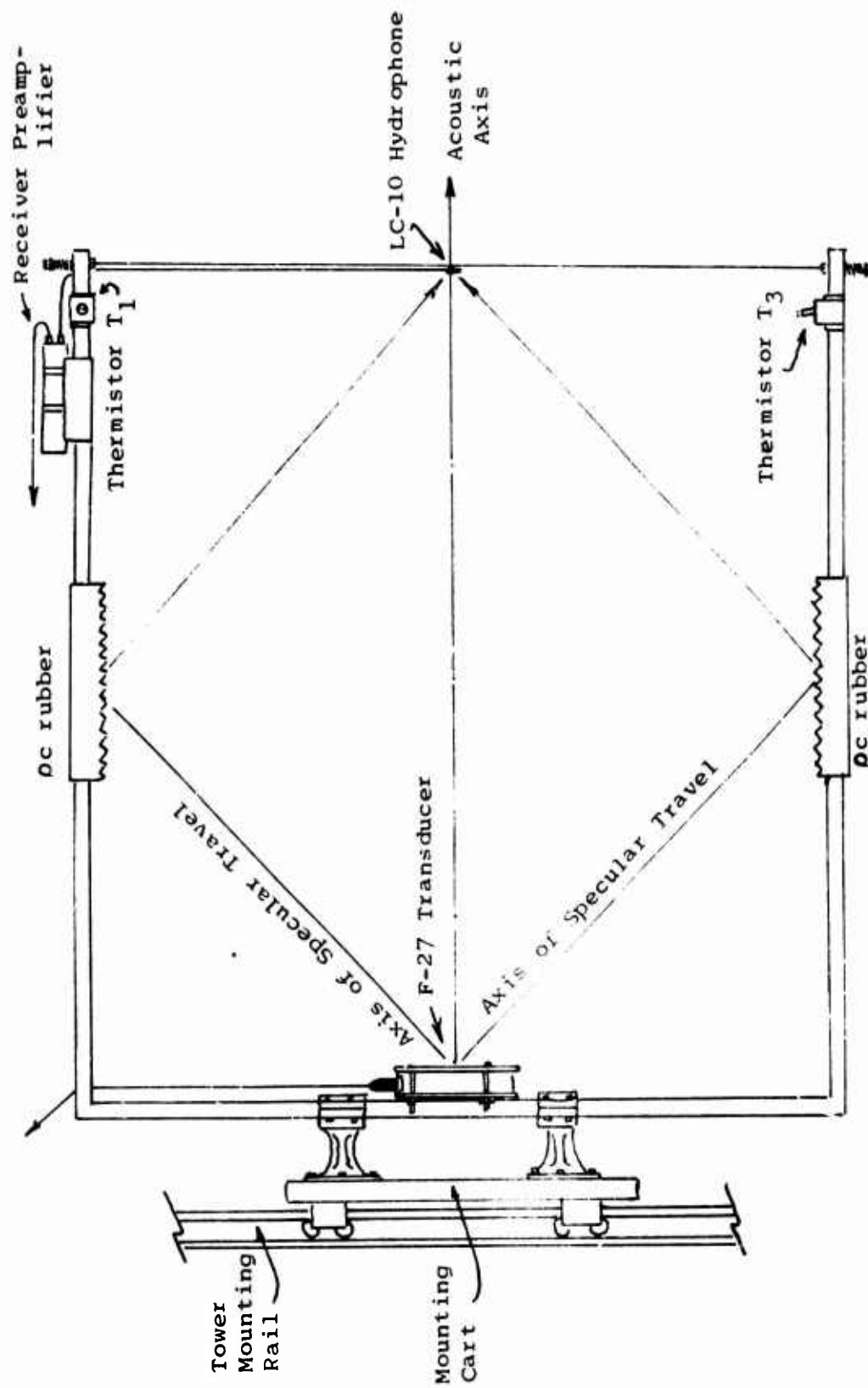
The apparatus (see Figure II-2) as designed was a square of height six feet and length two meters from the face of the transducer to the receiving hydrophone.

Three sides of the mount were of two inch diameter steel pipe and the fourth was one-eighth inch steel wire stretched by means of steel springs to a tension of one hundred and fifty pounds.

The square shape offered two distinct advantages, the shortness of the horizontal members minimized vibratory "tuning fork" effects in conjunction with the tension in the wire; it also minimized the number of reflection paths from source to receiver. Acoustic absorbent rubber (SoAB) mounted for eighteen inches along the upper and lower members also reduced possible reflections to the point that reflections were down 38 db relative to the signal when a pulsed 60 kHz signal was propagated.

An eleven inch steel disc was welded to the vertical pipe member directed perpendicularly to the sound axis as a mount for the transducer.

The completed design combined two necessary characteristics, rigidity and minimum self-generated hydrodynamic turbulence. The end of the apparatus holding the transducer was necessarily massive to insure rigidity about the propagating transducer.



ACOUSTIC APPARATUS IN EXPERIMENTAL CONFIGURATION

Figure II-2

Hydrodynamic turbulence scattered from the transducer was kept to a minimum by having the sound source end of the entire apparatus face the mean current.

The wire served as the mounting for the receiving hydrophone. Since the hydrophone was quite light in weight, the wire was rigid enough, while the small size of the wire precluded any significant hydrodynamic turbulence.

#### 5. The Transducer

The transducer utilized was the USRD type F-27. The transmitting array consists of fifty-five one-inch diameter lead metaniobate discs 0.22 inches thick, each of which is cemented to a one-inch diameter,  $\frac{1}{2}$  inch thick tungsten backing plate which acts approximately as a circular piston.

The F-27 transducer was designed for wide range unidirectional underwater sound projection at a frequency range of one kilohertz to forty kilohertz. The USRD recommends a limit of forty kilohertz since the beam pattern was considered to be too sharp at higher frequencies. This characteristic was advantageous to the experiment at sixty kilohertz and results were indicative of proper transducer selection, with the one-half power points measured at three degrees off the sound axis, and the side lobes at sixty degrees down forty db from the axial sound pressure level.

#### 6. The Receiving Hydrophone

The experiment required a small, sensitive hydrophone. The hydrophone used was the Atlantic Research Corp. type LC-10 with a diameter of 0.635 centimeters. For the sixty kilohertz carrier wave, the wavelength is 2.50 cm, making the wavelength 3.95 times

the diameter of the hydrophone. The sensitivity of the LC-10 is -110 dB re 1 volt per microbar at 60 kHz. The hydrophone was mounted vertically along the wire, with the central part of the cylinder perpendicular to the sound axis.

## B. PROCESSING ELECTRONICS

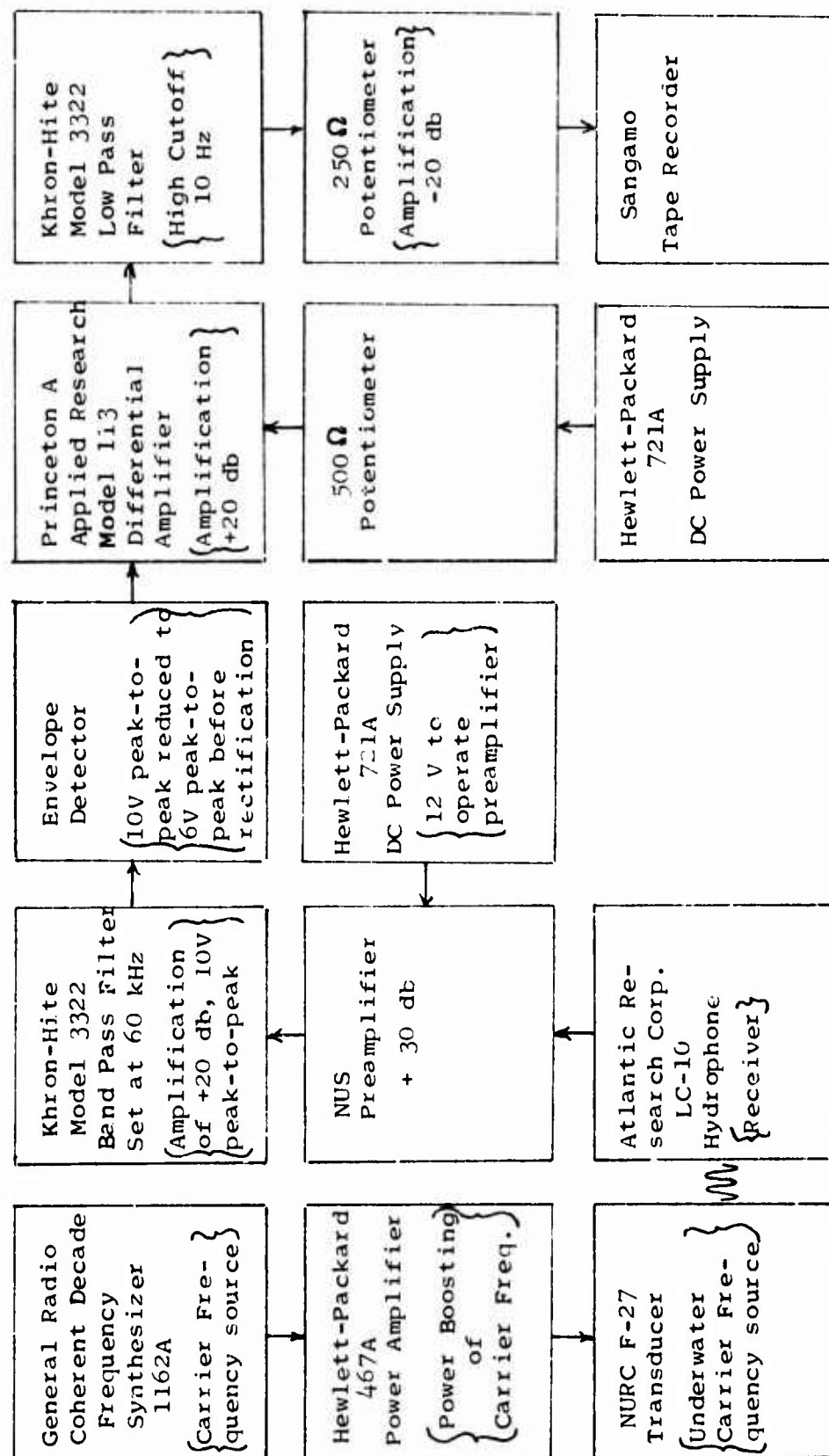
### 1. Overall System Operation

The processing electronics were designed to present the waveform of the amplitude modulation taking place on the acoustic carrier wave over a dynamic voltage range compatible with that of the recording equipment. Generally, it was desired to present a ten volt peak-to-peak signal at the entry leads of the envelope detector, and then process the waveform to view only the amplitude modulation. (See Figure II-3.)

The 60 kHz wave was originated in a GR 1162A frequency synthesizer which produced a one volt to four volt peak-to-peak waveform accurate and constant to within one millivolt over extended periods of operation.

The carrier waveform was then amplified to the voltage necessary to produce a ten volt peak-to-peak waveform at the envelope detector by an HP 467A power amplifier set at a fixed "times ten" setting. It was found to be advantageous to set the power amplifier at a fixed output and vary the output of the GR 1162A frequency synthesizer in order to achieve the desired voltage value at the envelope detector.

The power-amplified signal was then transmitted through one hundred and fifty feet of waterproof cable to the F-27 transducer and transmitted to the LC-10 receiving hydrophone.



ELECTRONICS PROCESSING AND DATA GATHERING SYSTEM CONFIGURATION

Figure II-3

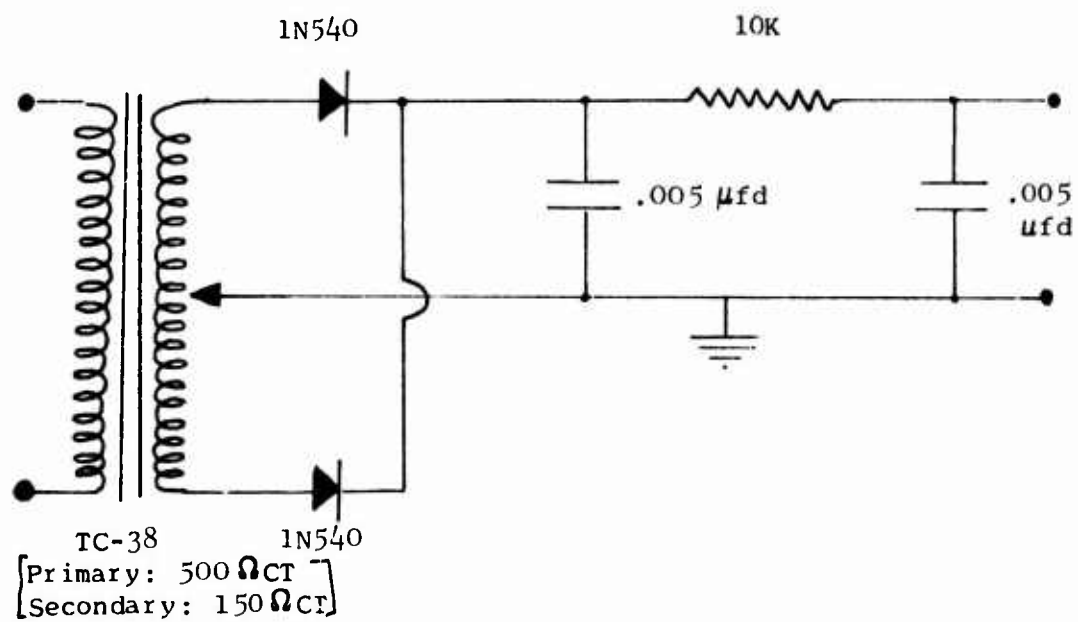


The signal as received by the LC-10 hydrophone was amplified by a fixed +30 db ( $\times 31.6$ ) gain NUS model 2010-030 voltage preamplifier in a watertight can mounted on the upper horizontal member of the mounting apparatus. The preamplifier required a twelve volt DC input, supplied by a Hewlett-Packard 721A DC power supply. The amplified signal was transmitted through one hundred and fifty feet of watertight cable to the initial electronic processing equipment.

The initial processing of the modulated waveform consisted of bandpass filtering the signal with a center frequency of 60 kHz through a pass band of  $\pm$  three hundred Hertz, with -24 db per octave attenuation on either side of the pass band. The unit utilized was the Khron-Hite model 3322 variable filter. In order to arrive at a ten volt peak-to-peak waveform at the output of the filter, a +20 db internal amplification feature of the filter was utilized. The output at this point consisted of a ten volt maximum peak-to-peak voltage with a low frequency environmental amplitude modulation superimposed upon it, varying to a maximum of sixteen per cent of the peak wave form.

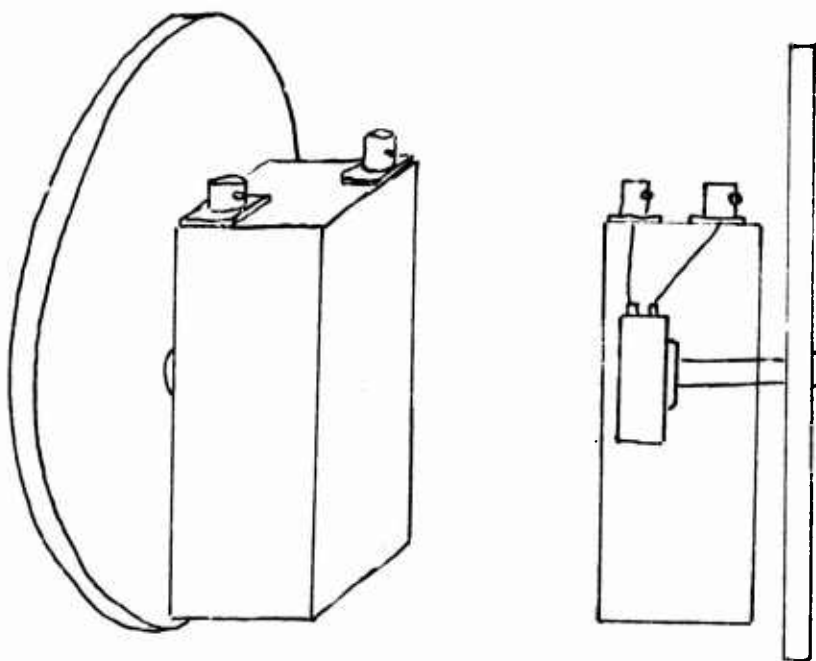
The above waveform was processed to detect the modulation by the envelope detector (see Figure II-4) described in detail in the following section. The salient point is that the ten volt signal was reduced by a factor of 0.6 in the audio transformer of the envelope detector.

The output of the envelope detector was the varying voltage induced by the amplitude modulation superimposed upon a small amount of DC voltage.



CIRCUIT DIAGRAM OF THE ENVELOPE DETECTOR

Figure II-4



DC COMPENSATION VOLTAGE POTENTIOMETER

Figure II-5

The envelope detector output was led into the positive input jack of a PAR Type 113 differential amplifier. The differential amplifier was used to eliminate the DC component of the amplitude modulation waveform. This was accomplished by simultaneously placing a DC voltage into the negative input jack. The DC voltage was supplied by an HP 721A DC power supply set at an output of four volts and reduced by means of a 500 ohm potentiometer to that (negative) voltage required to reduce the amplitude modulation to a purely alternating signal.

The potentiometer was necessary to assure a constant DC input level. The power supply/potentiometer combination was tested over a continuous two hour period with less than a one millivolt variation during the entire period. The potentiometer was configured with an outsized tuning wheel to allow minute (less than one millivolt) adjustments of the DC compensating voltage. (See Figure II-5.)

The compensated amplitude modulated signal was amplified by a factor of ten in the differential amplifier which made the varying voltage too high for the tape recorder, but which was subsequently reduced by a potentiometer. The ten-times setting was the minimum amplitude setting on the PAR 113.

The signal was subsequently passed through another Khron-Hite 3322 filter set in a four-pole Butterworth low pass configuration with a high cutoff of ten Hertz and a minimum frequency of 0.001 Hertz. The attenuation at the high frequency end was -48 db per octave. The internal amplification feature was not utilized.

The lowpass-filtered amplitude modulation signal was reduced to allow dynamic voltage range compatibility with the tape recorder. The reduction was accomplished by the use of a 250 ohm potentiometer set to reduce the low-pass filter output by a factor of ten. This reduction had the desirable result of restoring the amplitude modulation to its value relative to the six volt output of the audio transformer in the envelope detector.

The signal was recorded on a Sangamo model 3500 fourteen track tape recorder simultaneously with the oceanographic measurements during each run. The dynamic range of the tape recorder in the FM mode utilized was 0.01 to 2.50 volts, with all the amplitude modulation observed falling within this range.

The electronic processing design proved to be quite satisfactory except for details in various components which were amply compensated for during the course of the experiment.

## 2. The Envelope Detector

The envelope detector was adapted from an ocean wave height detector and proved to be satisfactory. The detector is essentially a double rectification device with an RC filter and a capacitive lead to ground potential to discharge DC voltages built up in the silicon diode (1N540) rectifying elements. (See Figure II-4.)

The time constant for the RC filter at a carrier frequency of 60 kHz is fifty microseconds and with double rectification there was no appreciable decay between peaks.

The envelope detector proved sensitive to acoustic modulation of as little as one twentieth of one per cent in the laboratory tanks.

Possible improvements to the detector would be the substitution of larger capacitors in the filtering element to reduce the substantial (1.3 volt) DC component in the output.

### III. EXPERIMENTAL PROCEDURE

#### A. GENERAL

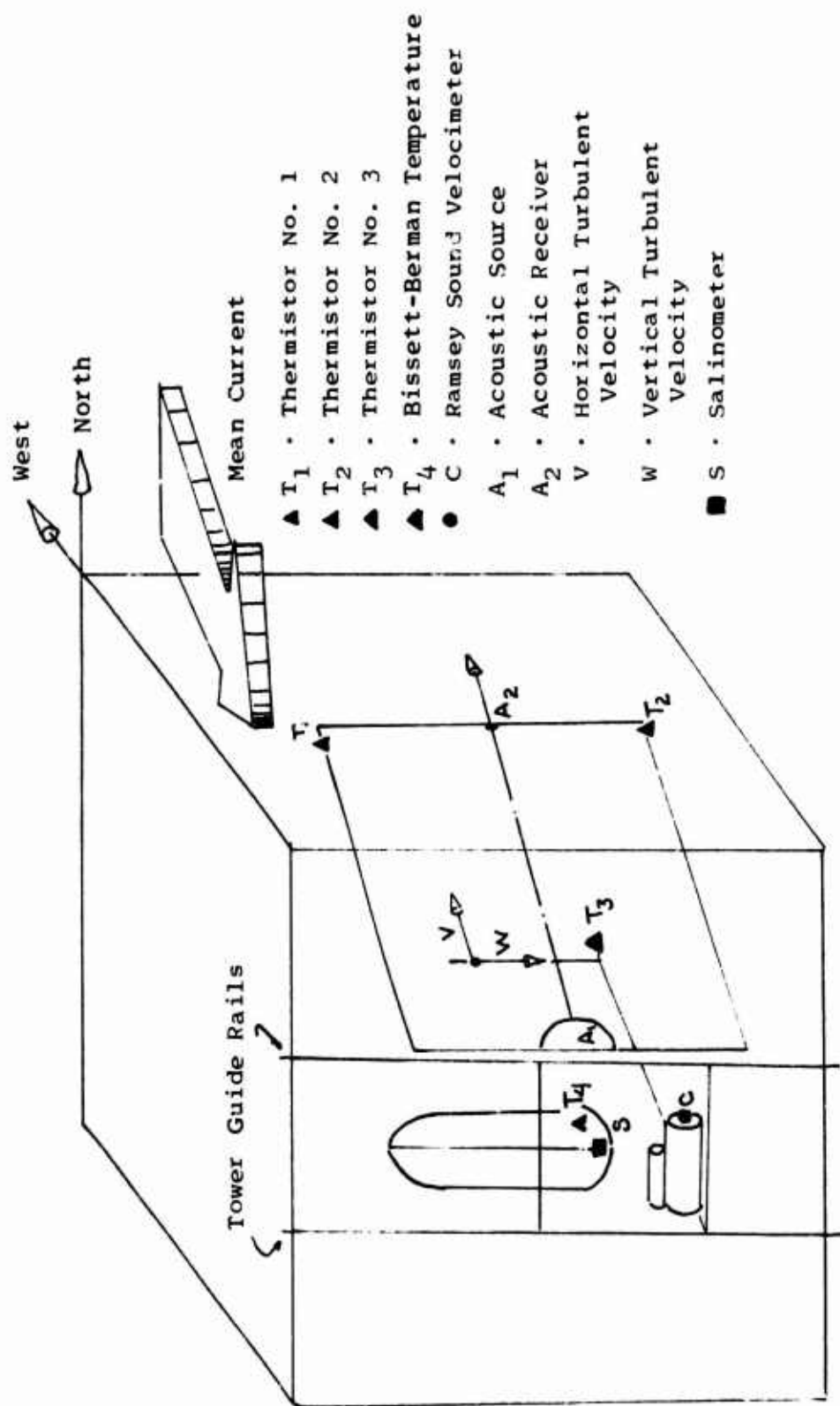
The amplitude modulation runs were performed on the twenty-first of October 1971 at the NUC Oceanographic Research Tower at San Diego. The runs were twenty minutes in length, to allow enough time for major low frequency ocean phenomena to be adequately described and measured.

The ocean sensors were mounted around the acoustic sensors in order to define the volume of ocean over which the amplitude modulation took place. (See Figure III-1.) The sensors were selected to measure temporal variations of temperature, salinity, turbulent velocity, wave height and the speed of sound during the period of acoustic measurement. During each run, the oceanographers present also measured and observed weather and sea surface conditions as background for the more precise information presented by the ocean sensors. The ocean sensors used are listed below with respect to the parameters they measured.

##### 1. Temperature

a. Three thermistors located about the apparatus obtained temporal variation of temperature in the volume of water under observation.

b. One Bisset-Berman temperature sensor measured absolute values of temperature near the F-27 transducer over a range of  $14^{\circ}\text{C}$  to  $19^{\circ}\text{C}$ .



SPATIAL CONFIGURATION OF ACOUSTIC AND OCEAN SENSORS IN RELATION TO WATER VOLUME, MEAN CURRENT AND CARDINAL BEARINGS (LESS SOME DETAIL)

Figure III-1

## 2. Salinity

One Bisset-Berman salinometer located near the F-27 transducer measured temporal variation in salinity over a range of 37.5 to 39.5 parts per thousand.

## 3. Particle Velocity

An electromagnetic flowmeter measured one component, (u) or (v), of horizontal velocity and the vertical component (w). The flowmeter was located about one-half of the distance between the acoustic sensors 1.13 meters downstream from the acoustic axis.

## 4. Speed of Sound

The sound speed was measured as an absolute value with temporal variations by a Ramsey Corp. Mark I SVTD probe located near the base of the acoustic apparatus.

## 5. Wave Height

The wave height was measured by two sensors, a pressure wave height indicator and a Baylor gauge.

As noted above, not all the ocean sensors were operating during all runs. Five runs were made, with runs two through four considered suitable for analysis. The general scheme was to perform two runs near the surface, the first run with the electromagnetic flowmeter measuring in the u and v (horizontal) directions, and the second with the flowmeter measuring in the u and w (or vertical) directions. Three runs followed at deeper depths with the flowmeter in the u-w configuration. The first run was aborted when the flowmeter malfunctioned, but runs two through five proved to be successful. A listing of the runs taken is presented below; all times are local and the date is 21 October



1971. Mean water level was 6.7 meters below the concrete deck of the NUC tower. The depths are in meters below mean water.

Run 1.

depth: 4.3 meters,

times: start: 1419 stop: 1439

comment: The data from this run were not analyzed due to a malfunction of the electromagnetic flowmeter.

Run 2.

depth: 4.3 meters,

times: start: 1530 stop: 1550

sensors: turbulence: 1. u (horizontal) velocity

2. w (vertical) velocity

3. sound velocity

4. Both wave heights

temperature: 1. thermistor  $T_3$

2. Bissett-Berman temperature

salinity: 1. Bissett-Berman salinometer

Run 3.

depth: 9.3 meters,

times: start: 1616 stop: 1636

sensors: turbulence: 1. u velocity

2. w velocity

3. sound velocity

4. both wave heights

temperatures: 1. thermistors  $T_1$ ,  $T_2$ , and  $T_3$

2. Bissett-Berman temperature

salinity: Bissett-Berman salinometer

Run 4.

depth: 14.3 meters,

times: start: 1648 stop: 1708

sensors: all sensors operating, see sensor information  
for Run 3.

Run 5.

depth: 7.3 meters,

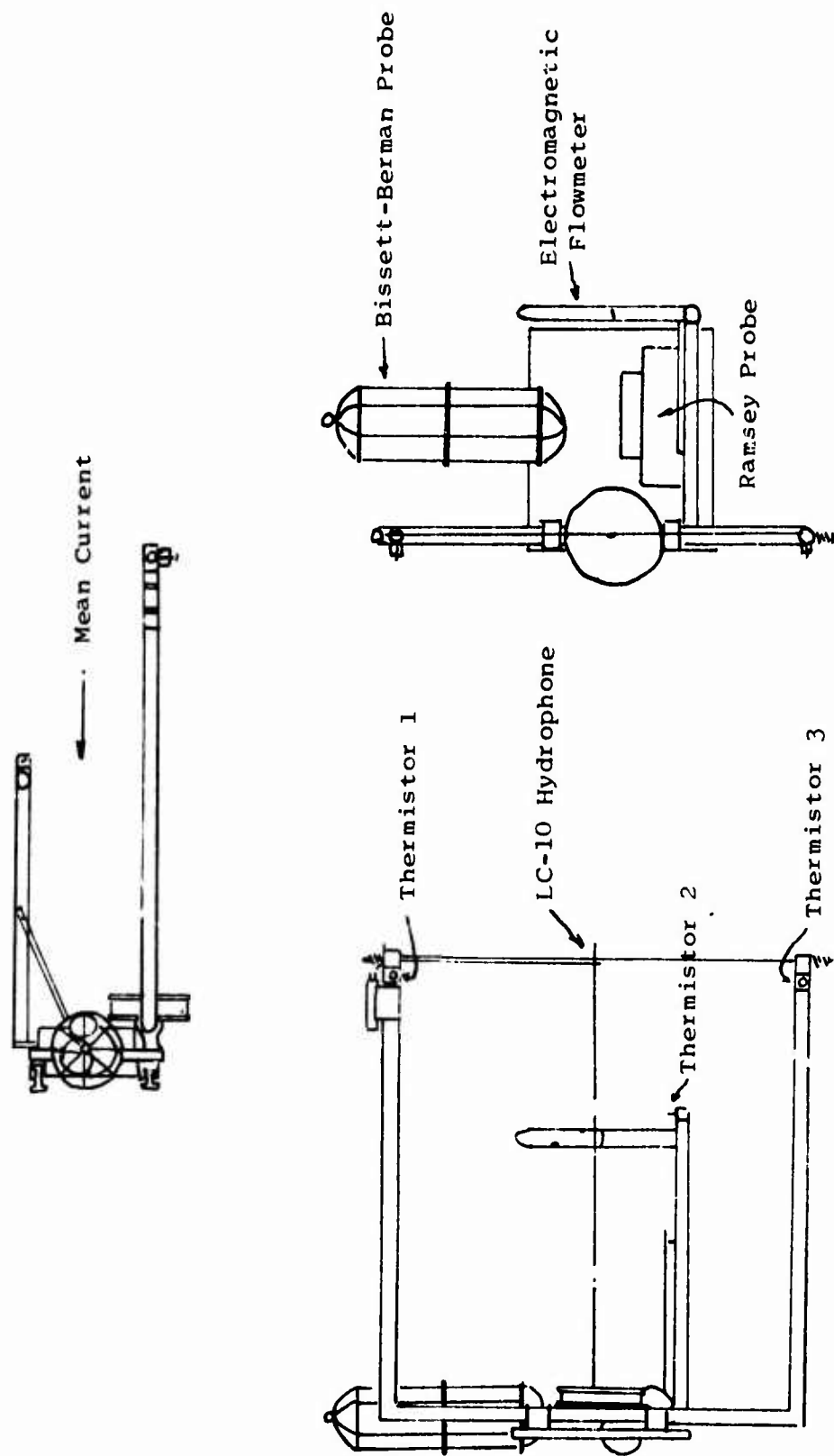
times: start: 1728 stop: 1748

sensors: all sensors operating, see sensor information  
for Run 3.

#### B. CONFIGURATION OF THE OCEAN SENSORS IN RELATION TO THE ACOUSTIC APPARATUS

Two factors determined the configuration of the ocean sensors relative to the acoustic apparatus. First, the volume of water around the acoustic axis had to be described adequately for temporal variations, and second, the ocean sensors had to be placed such that there was little or no interference from downstream turbulence from the sensors themselves.

Temperature variations and turbulent velocities were considered to be the most important parameters affecting the amplitude modulation. Temperature variations were recorded by the three thermistors and the Bissett-Berman temperature probe. The Bissett-Berman probe was placed centrally between the acoustic apparatus and the mount for the flowmeter due to its size. Two of the thermistors were placed above and below the receiving hydrophone in the vertical plane of the acoustic axis, while the third was placed along the horizontal plane of the axis midway between the transducer and the hydrophone. (See Figure III-2.)



CONFIGURATION OF ACOUSTIC AND OCEAN SENSORS DURING THE  
EXPERIMENTAL RUNS

Figure III-2

The electromagnetic flowmeter was placed so that it measured the turbulence occurring 113 centimeters from the center of the acoustic axis.

All other sensors were integral parts of large instruments and were placed between the acoustic apparatus and the mount for the flowmeter to minimize the effect of their wake turbulence.

Figure III-3 shows the distances between various sensors. The mean current was as shown in Figure III-1, with the apparatus up-stream of all other equipment.

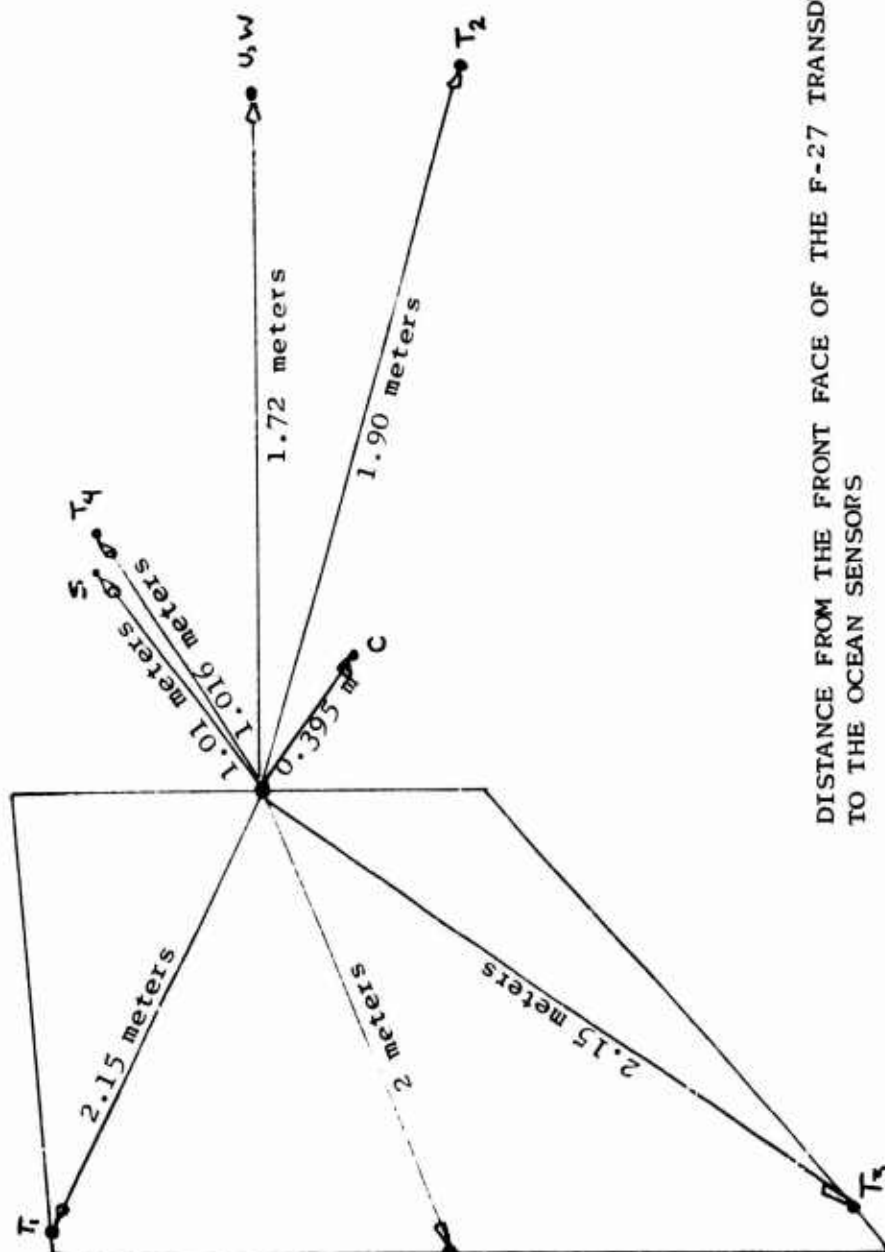
#### C. DETERMINATION OF RUN DEPTHS

The mean water depth at the NUC tower was 16.4 meters (54 feet) during the time of the runs. Four depths were selected in order to observe near-surface to near-bottom amplitude modulation effects. The first two runs were at an estimated water depth of 4.3 meters, the third at an estimated 9.3 meters, the fourth (and deepest) at 14.3 meters, and the last run was at 7.3 meters.

These depths were chosen in order to cover the entire water column and to minimize interference from the underwater beam structure of the tower. The shallowest depth was far enough from the surface to avoid undue stresses on the apparatus, while the deepest depth was far enough away from the bottom to avoid inadvertent contact.

#### D. THE NUC TOWER

The NUC tower is located approximately one mile off Mission Beach, California in 16 meters of water. The apparatus as depicted in Figure III-2 was mounted on the west side of the tower.



DISTANCE FROM THE FRONT FACE OF THE F-27 TRANSDUCER  
TO THE OCEAN SENSORS

Figure III-3

The electronics were installed on the main deck, while the apparatus was handled from the second deck immediately below. (See Figure III-4.)

Environmental factors were observed at the tower while the runs were in progress. The factors noted were:

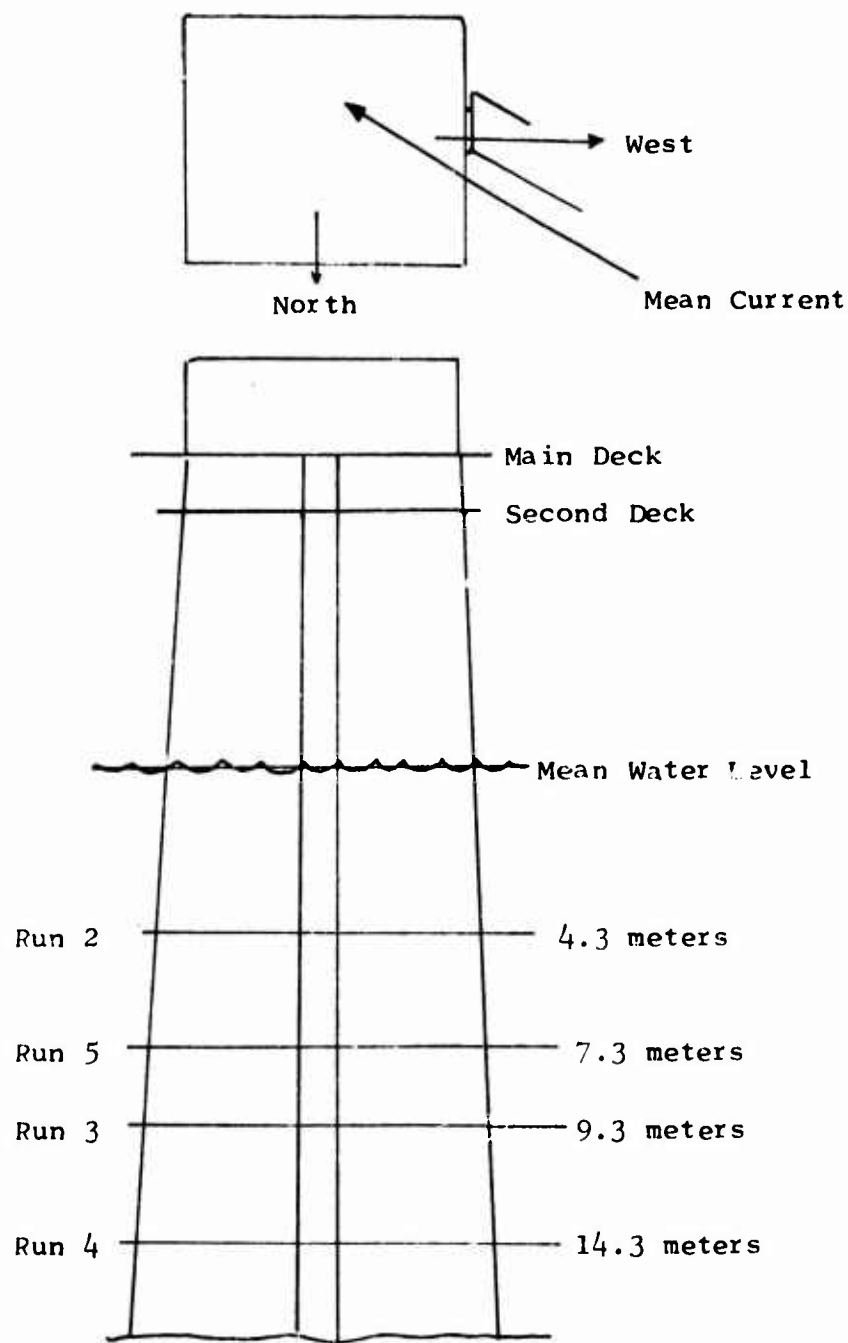
1. swell: from the northwest, two feet from crest to trough, with an estimated period of four to five seconds.
2. wind:  $310^{\circ}$ ,  $300^{\circ}$ ,  $320^{\circ}$  from 9 knots to 7.5 knots during the course of the experiment
3. wind waves six inches from crest to trough
4. some bubbles visible on the surface

#### E. UNUSUAL EFFECTS

Unusual effects were those effects arising from considerations other than physical parameters and naturally occurring bubble populations. All the unusual effects noted were due to animal and plant life in and around the apparatus during each run.

The area around the tower supported a large population of small fish ranging up to eight inches in length. It was noted that fish were attracted to the apparatus when it was lowered initially, but resumed their normal habits after a short period of time. Fish could not affect the wire mounted hydrophone, nor were any effects noted, since fish swimming through the sound beam would affect the modulation for a minimal amount of time.

Plant life could affect the equipment by drifting into it. There was a small amount of kelp in the area at all depths, but no acoustic effects noted during the runs.



THE NUC TOWER: DEPTHS OF EACH RUN,  
AND ORIENTATION

Figure III-4

Boat traffic around the tower had some effect, and could have been sources of turbulent velocity and bubbles which could noticeably affect the amplitude modulation of the acoustic signal. This effect was quite apparent during the Run 5, when there was a boat idling in the area.

#### F. ORGANIZATION OF INDIVIDUAL RUNS

All information from each sensor was recorded on Sangamo Model 3500 tape recorder. Between individual runs, all sensors were monitored for proper output. After a depth change, a few minutes were allowed for all instruments to be checked for proper operation. When all instruments were checked and operating, the tape recorder was started. The simultaneous start and stop times controlled by the tape recorder insured an accurate run period for all sensors.

The recorder channel allocation was as follows:

- Channel 1.: Ramsey probe temperature
- Channel 2.: Ramsey probe velocimeter
- Channel 3.: Bissett-Berman salinometer
- Channel 4.: Turbulent velocity (v)
- Channel 5.: Turbulent velocity (w)
- Channel 6.: Baylor wave height probe
- Channel 7.: Amplitude modulation
- Channel 8.: Pressure wave height probe
- Channel 9.: Thermistor  $T_1$
- Channel 10.: Bissett-Berman temperature
- Channel 11.: Thermistor  $T_2$



Channel 12.: Thermistor  $T_3$

Channel 13.: Phase modulation

Channel 14.: Voice

#### IV. ANALYSIS OF ACOUSTIC RESULTS

##### A. INITIAL DATA REDUCTION

###### 1. General

The recorded analog data consisted of four twenty minute runs at varying depths. The dynamic range of the recorded voltage was 0.01 to 1.10 volts. The data reduction scheme involved three steps; transfer of the analog data from magnetic tape to a Brush Mark 200 strip-chart recorder, conversion of the printed analog data to digitized data and computer analysis of the digitized data.

###### 2. Magnetic Tape to Strip-Chart Record

The tape recorded data was originally collected at a tape speed of  $1 \frac{7}{8}$  inches per second. When replayed for the Brush Mark 200 eight channel strip-chart recorder, the data was played at  $7\frac{1}{2}$  inches per second, making the strip-chart data for each run one fourth as long as the real-time runs. This had the effect of compressing a twenty minute run to five minutes of strip-chart readout and increasing the apparent frequencies of modulation by a factor of four. At a strip-chart recorder speed of five millimeters per second, each printed run was presented on 1.5 meters of chart.

The strip-chart recorder was set to present the data at 50 millivolts per millimeter, or 1.27 volts per inch.

The other seven strip-chart channels were used to visually compare some of the ocean sensor outputs to the output from the amplitude modulation data. The other seven strip-chart channels

were allocated as follows:

Channel 1.: Salinometer

Channel 2.: v velocity

Channel 3.: w velocity

Channel 4.: Amplitude modulation

Channel 5.: Baylor wave height

Channel 6.: Thermistor  $T_3$

Channel 7.: Bissett-Berman temperature

Channel 8.: Velocimeter

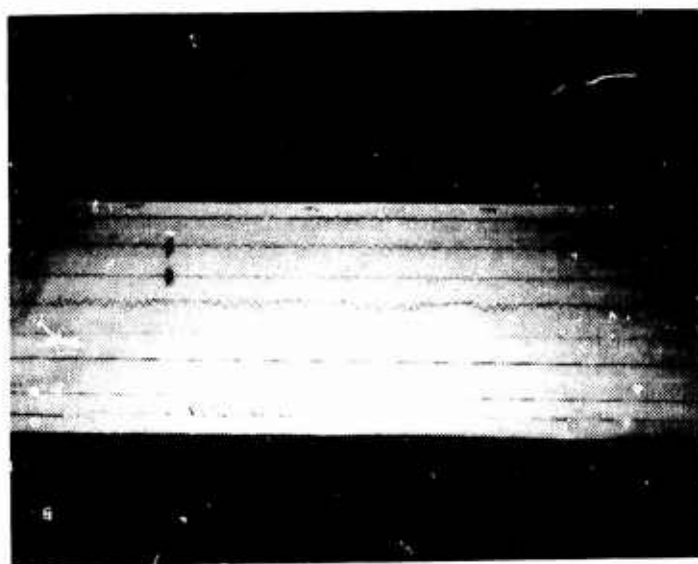
It was then possible to make preliminary visual comparison of the waveforms of the recorded outputs in order to observe any similarities in the waveforms. (See Figures IV-1 through IV-4.)

### 3. Strip-Chart Record to Digital Tape

The strip-chart records of each of the four runs to be analyzed were taken to the Fleet Numerical Weather Central facility at Point Piños, California to be recorded as digitized data on magnetic tape. The data were digitized directly from the strip chart recording on the facility's tracing digitizer. The digitizer recorded data points in increments of hundredths of an inch in the y-direction for each one one-hundredth of an inch of travel in the x-direction, and transferred the x and y values to a nine-track magnetic tape.

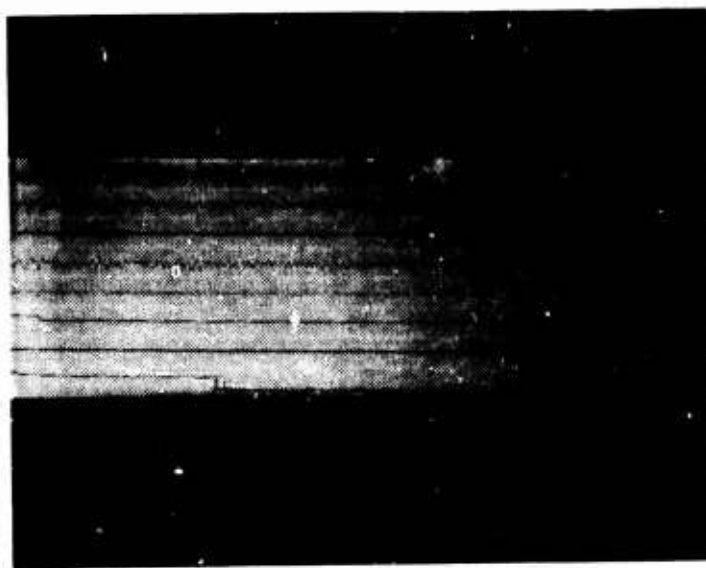
The digitizing was initiated at the start of each run, visible on the strip-chart as a vertical line produced when the tape recorder run was commenced. The zero amplitude value in the y-direction and the zero time value in the x-direction were taken at this line. There were data excursions in the positive and

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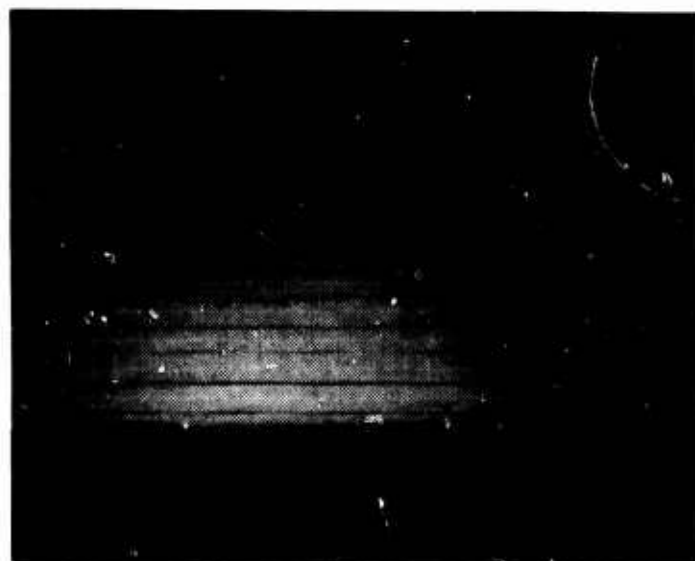
EIGHT CHANNEL BRUSH RECORDER READOUT FOR RUN 2, DEPTH: 4.3 meters. top to bottom: salinometer, v-turbulence, w-turbulence, amplitude modulation, wave height, thermistor 3, Bissett-Berman temperature and velocimeter readouts.

Figure IV-1



EIGHT CHANNEL BRUSH RECORDER READOUT FOR RUN 5 DEPTH: 7.3 meters. For channel identification, see Fig. IV-1.

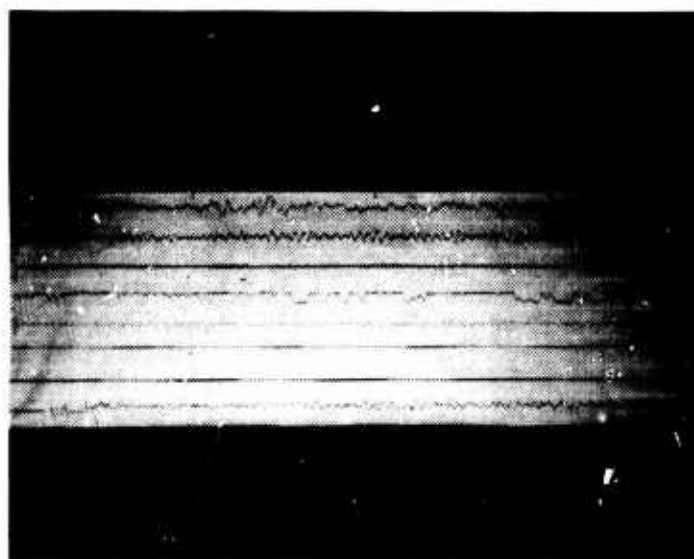
Figure IV-2



EIGHT CHANNEL BRUSH RECORDER READOUT FOR RUN 3 DEPTH:  
9.3 meters. For channel identification, see Fig. IV-1.

Figure IV-3

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best available copy.



EIGHT CHANNEL BRUSH RECORDER READOUT FOR RUN 4 DEPTH:  
14.3 meters. For channel identification see Fig. IV-1.

Figure IV-4

negative directions, measured in one-hundredths. The conversion factors were 0.0127 volts per data point in the y-direction, and 0.01265 seconds (real time) per data point in the x-direction.

#### 4. Digitized Data from Magnetic Tape to IBM Data Cards

The conversion of the digitized data from magnetic tape to data cards for analysis on the IBM 360 digital computer was accomplished in the Fleet Numerical Weather Central's CDC 5000 computer.

The procedure utilized Program I, and converted at least 5600 data points for each run, corresponding to nineteen minutes of real data collection time. This decrease in data time was due to an uncertainty of stopping points on the strip-chart readouts, and allowed an identical number of data points for each run.

#### B. ANALYSIS OF ACOUSTIC DATA

The data from each run were analyzed on the Naval Postgraduate School's IBM 360 digital computer.

The analyses were performed on time-varying amounts of fractional amplitude modulation. The raw data as presented on data cards was in varying values expressed as positive and negative hundredths of an inch. The conversion was common to all programs and is described below:

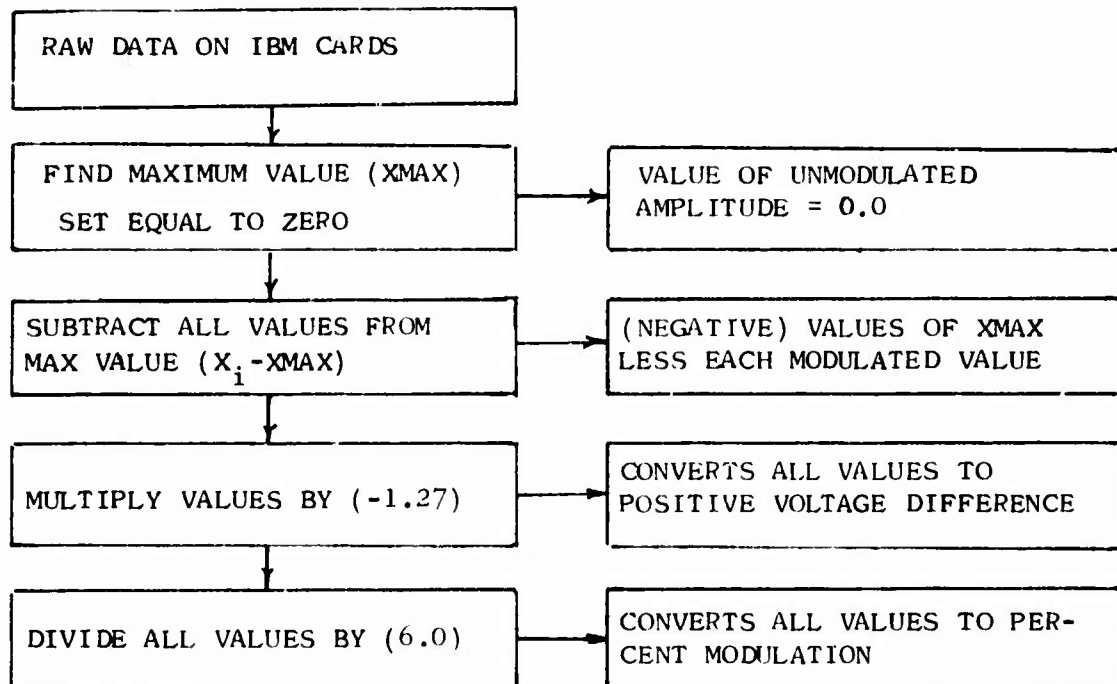
- a. Each run was scanned for the maximum positive value.
- b. The maximum positive value was subtracted from all other values and converted to 0.0. This level corresponded to the unmodulated amplitude. All data values were now negative values, all less than zero except for the maximum value.

c. Each value was multiplied by a conversion factor consisting of (-1.27) which converted the data points to positive voltage values, and (1/6) which made each point a fractional value of the maximum amplitude of six volts. The resultant data points could now be expressed as fractional modulation:

$$F = \left[ \frac{(\text{unmodulated voltage}) - (\text{modulated voltage})}{(\text{unmodulated voltage})} \right]$$

The converted raw data was then analyzed for the mean, variance, temporal correlation and power spectral density of each run in Program II. Program III computed a histogram of the amount of fractional modulation versus the number of occurrences of each value. An independent calculation of the mean, mean square and variance was made in Program IV. (See Figures IV-5 through IV-7.) The results of the various analyses are presented in the following section.

# DATA ANALYSIS SCHEME: PROGRAM II



AT THIS POINT, ALL VALUES MAY BE EXPRESSED AS:

$$F = \left[ \frac{(\text{UNMODULATED VOLTAGE}) - (\text{MODULATED VOLTAGE})}{(\text{UNMODULATED VOLTAGE})} \right]$$

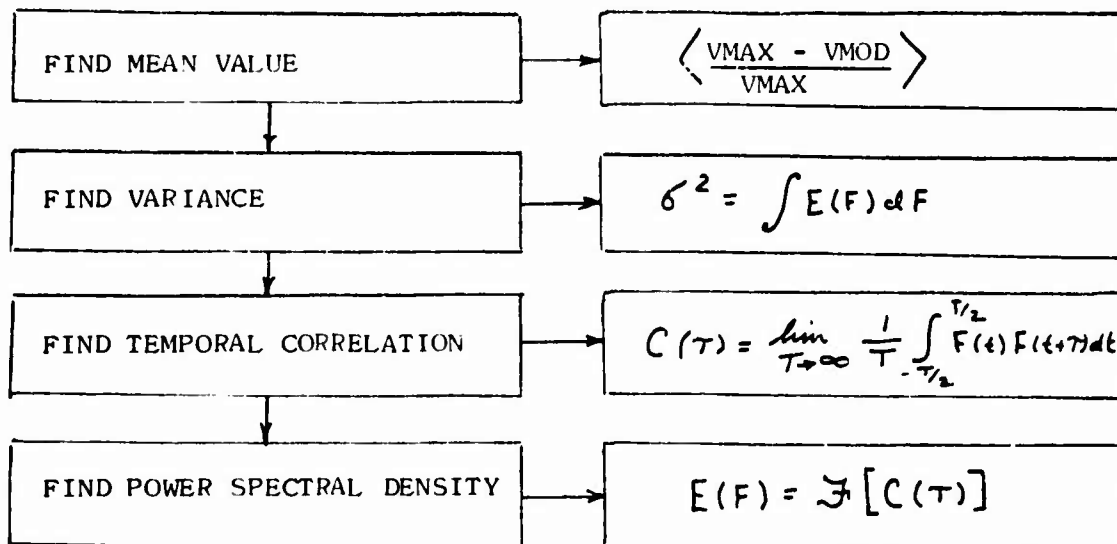


Figure IV-5



DATA ANALYSIS SCHEME: PROGRAM III

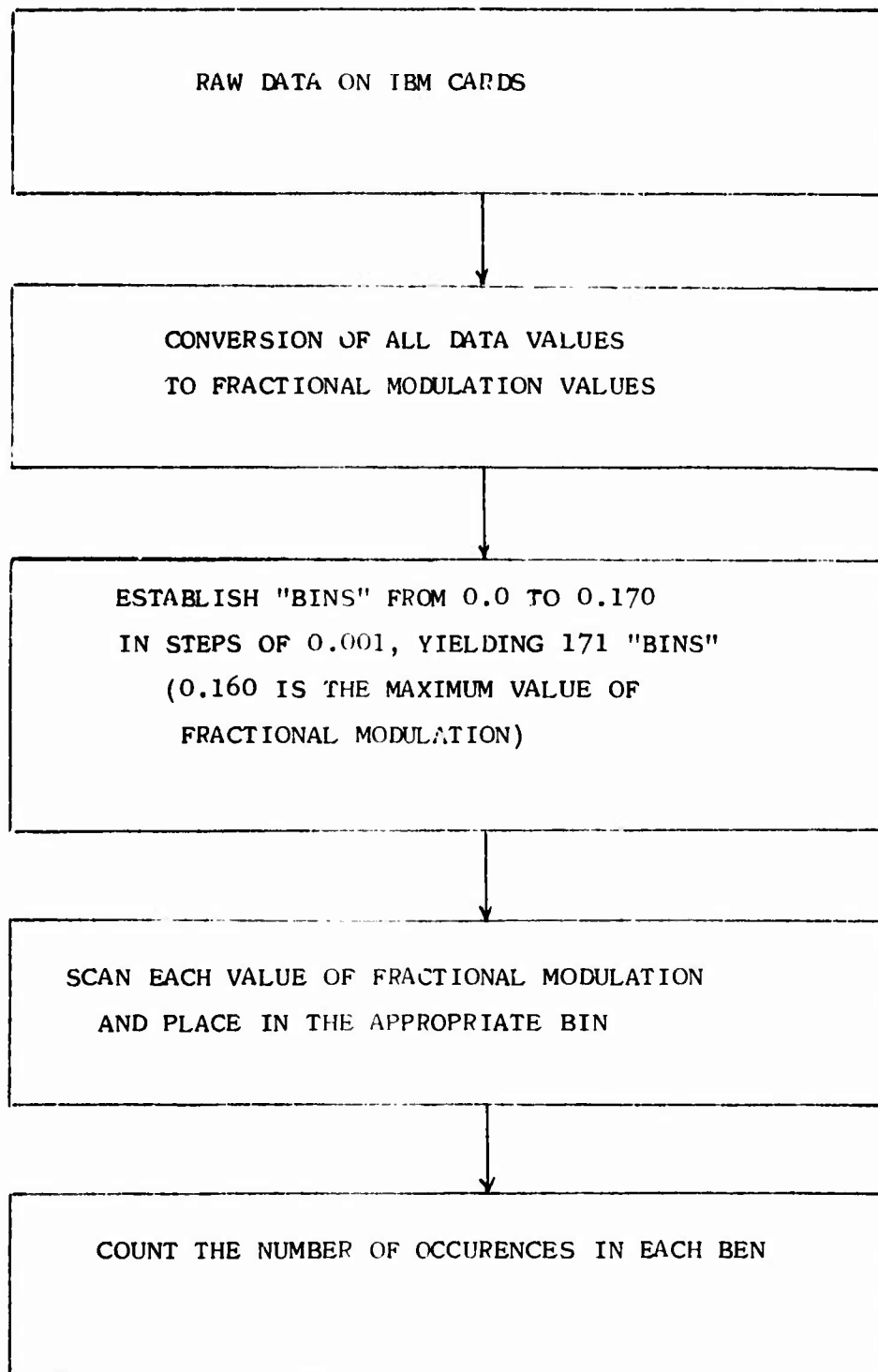


Figure IV-6

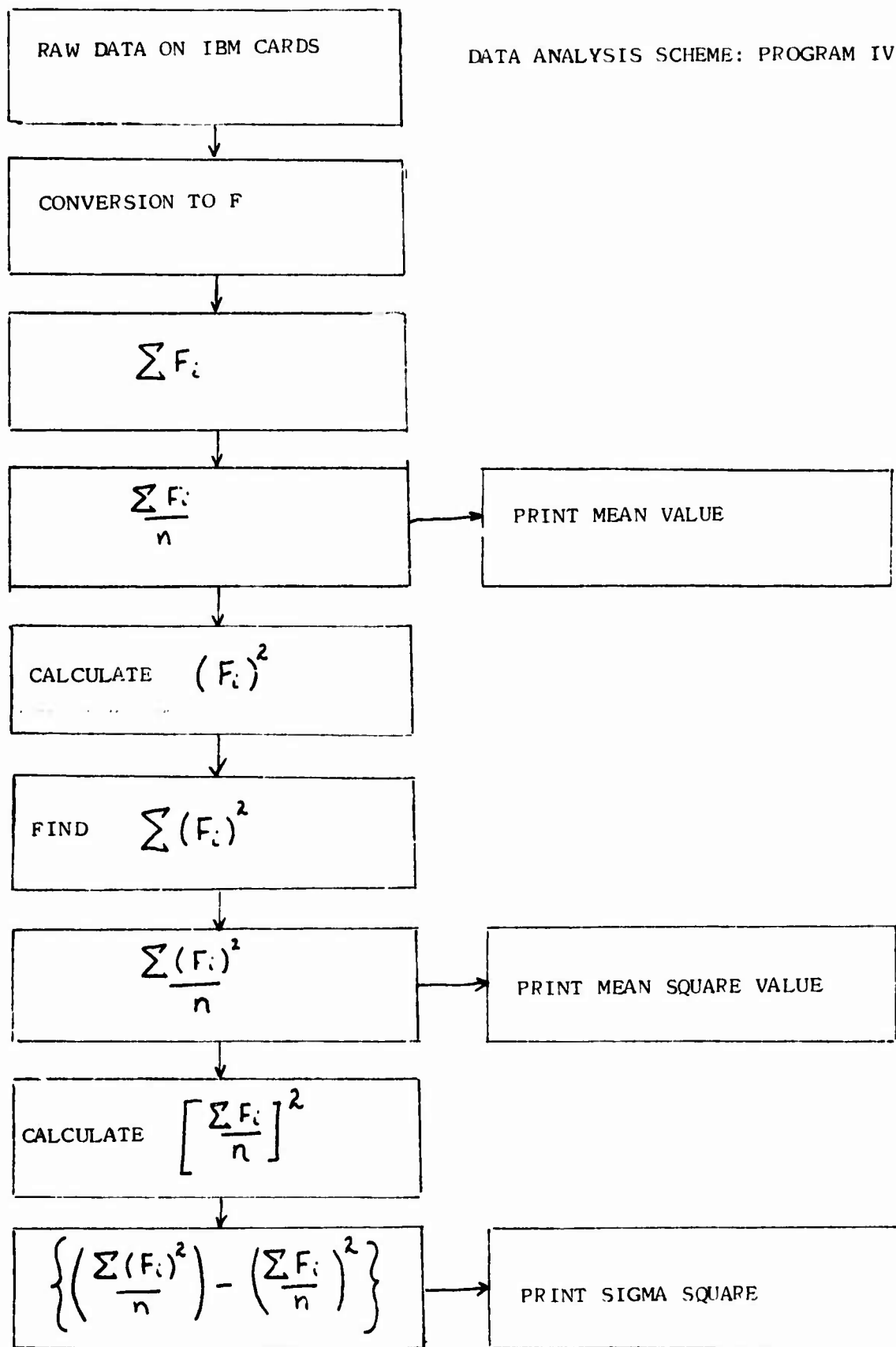


Figure IV-7

## V. ACOUSTIC RESULTS

### A THE STRIP-CHART READOUTS

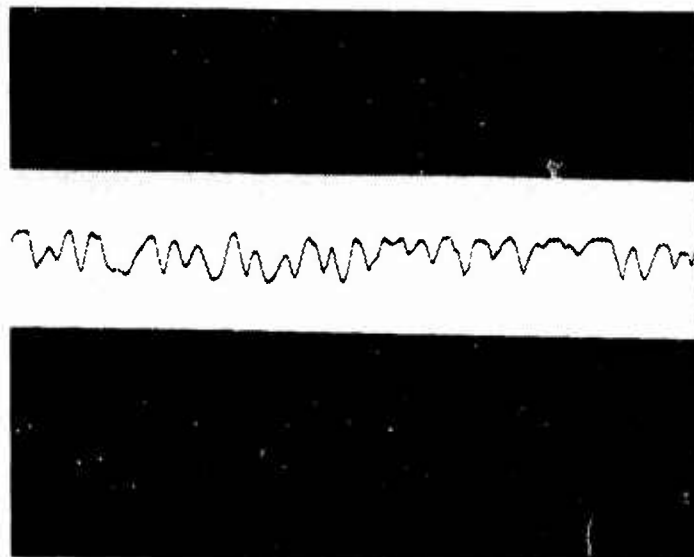
Figures V-1 through V-4 are representative 168-second portions of each twenty-minute run. The variation in voltage from peak-to-peak is approximately one volt, and the apparent frequencies of variation are four times the actual frequencies occurring during each run.

From visual inspection of the strip-charts, it may be seen that the modulations occurring during the two shallow runs (4.3 and 7.3 meters) were more active in frequency than during the two deeper runs (9.3 and 14.3 m.). It must be noted that the 7.3 meter run was conducted after the other runs, and that during most of the period a fifty-foot motor launch was in the area of the tower with propellers turning. The launch served as a bubble injection device as well as a turbulence generator. The run was not invalidated because the modulation effect of bubbles artificially generated during this run could be compared with that of the ambient medium of the previous runs.

### B. MEANS AND VARIANCES

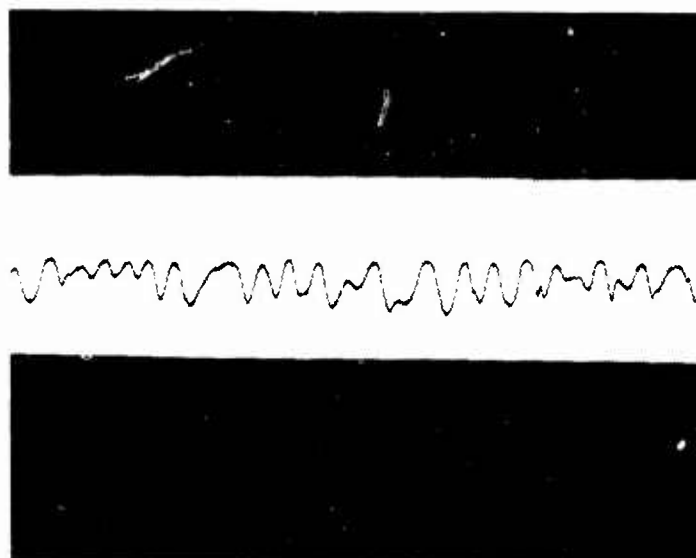
The mean and variance of the amplitude modulation for each run was determined by Computer Run IV. The formulation was based on the fractional modulation F, and was calculated as shown:

$$F = \left[ \frac{V_{MAX} - V_{MOD}}{V_{MAX}} \right]$$



BRUSH RECORDER READOUT OF AMPLITUDE MODULATION RUN TWO DEPTH:  
4.3 meters. vertical lines indicate four seconds in real time,  
horizontal lines indicate 0.25 volts.

Figure V-1



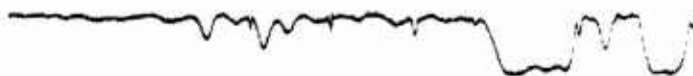
BRUSH RECORDER READOUT OF AMPLITUDE MODULATION RUN 5 DEPTH:  
7.3 meters. see caption for Fig. V-1.

Figure V-2



BRUSH RECORDER READOUT OF AMPLITUDE MODULATION RUN 3 DEPTH:  
9.3 meters. see caption for Fig. V-1.

Figure V-3



BRUSH RECORDER READOUT FOR AMPLITUDE MODULATION RUN 4 DEPTH:  
14.3 meters. see caption for Fig. V-1.

Figure V-4

which resulted in the time series  $F(t)$ ; then,

$$\langle F(t) \rangle = \frac{\sum F_i}{n}$$

where  $n = 5600$

and

$$\langle F^2(t) \rangle = \frac{\sum (F_i)^2}{n}$$

and

$$\sigma^2 = \{ \langle F^2(t) \rangle - \langle F(t) \rangle^2 \}$$

The results obtained from each run were compared with the results of Computer Run II, which determined the mean, temporal correlation and the power spectral density for each run. Effectively, for Program II the mean,  $\langle F(t) \rangle$ , was determined as in Program IV so that comparison was superfluous.

The variance,  $\sigma^2$ , can be formulated in three ways; first as shown above, secondly as the zero-time temporal correlation value,

$$\sigma^2 = C(\tau) \Big|_{\tau=0} = \lim_{T \rightarrow \infty} \frac{1}{T} \int_{-T/2}^{T/2} F(t) F(t+\tau) dt$$

and thirdly as the value of the area under the curve for the power spectral density:

$$\sigma^2 = \int E(F) dF$$

It should be noticed that at all depths, the three values of the variance are quite close. The slightly different values of  $\sigma^2$  from the calculation for  $C(0)$  and from the simple sorting calculation (Program IV) are solely due to round off error for the latter calculation.

The value of  $\sigma^2$  derived from the integral of the power spectral density has the largest numerical value. This is the result of limiting the higher frequencies from the calculation of the Fourier pass band, resulting in a rounded "window". This rounding off resulted in a very slightly larger area under the power spectral density curve than would otherwise occur.

The results for each run are shown below:

Run 2: 4.3 meters

$$\text{Program IV: } \langle F \rangle = 0.0526$$

$$\langle F^2 \rangle = 0.0041$$

$$\sigma^2 = 0.0013$$

$$\text{Program II: } \langle F \rangle = 0.05261$$

$$\sigma^2 = C(0) = 0.00131$$

$$\sigma^2 = \int E(F) dF = 0.0014469$$

Run 5: 7.3 meters

$$\text{Program IV: } \langle F \rangle = 0.0460$$

$$\langle F^2 \rangle = 0.0031$$

$$\sigma^2 = 0.0010$$

$$\text{Program II: } \langle F \rangle = 0.04604$$

$$\sigma^2 = C(0) = 0.00096$$

$$\sigma^2 = \int E(F) dF = 0.0009756$$

Run 3: 9.3 meters

$$\text{Program IV: } \langle F \rangle = 0.0303$$

$$\langle F^2 \rangle = 0.0016$$

$$\sigma^2 = 0.0007$$

$$\text{Program II: } \langle F \rangle = 0.03026$$

$$\sigma^2 = C(0) = 0.00067$$

$$\sigma^2 = \int E(F) dF = 0.0007378$$

Run 4: 14.3 meters

$$\text{Program IV: } \langle F \rangle = 0.0470$$

$$\langle F^2 \rangle = 0.0040$$

$$\sigma^2 = 0.0018$$

$$\text{Program II: } \langle F \rangle = 0.0472$$

$$\sigma^2 = C(0) = 0.00173$$

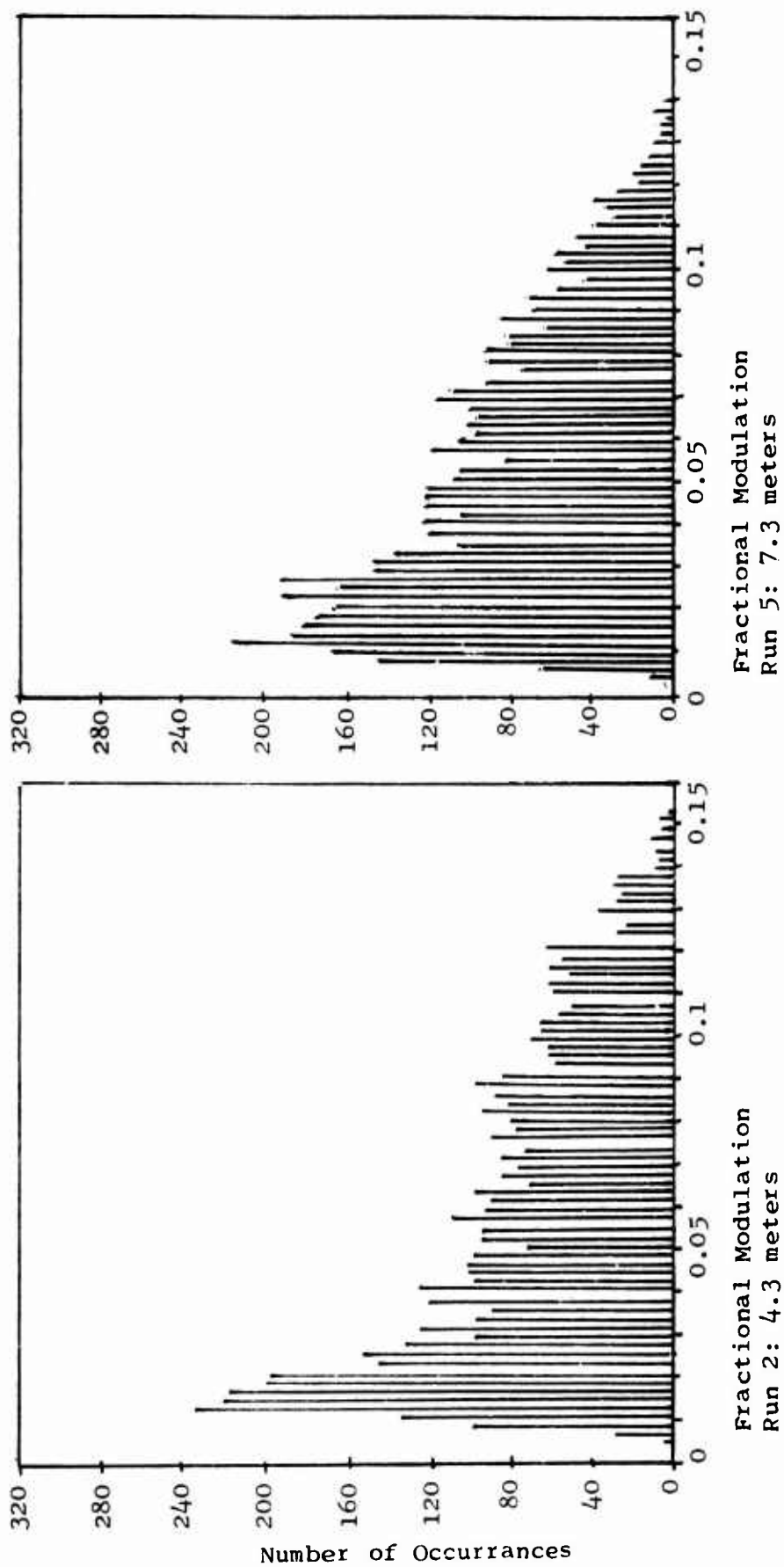
$$\sigma^2 = \int E(F) dF = 0.0019268$$

The probability density of fractional modulation values for each run is shown on the histograms. (Figures V-5 through V-7.) Although the mean values calculated above do not differ in any orderly manner with depths, the histograms (as well as Figures V-1,2,3 and 4) show that for the shallow depths there are greater numbers of short duration, high modulation values, whereas for the greater depths, the distribution of fractional modulation centers heavily about the lower values. The bimodal P.D.F. at the greatest depth (14.3 meters) shows that onset of infrequent occurrences of very high modulation values. This is the statistical representation of the type of modulation observed on Figure V-4, where the modulation assumes a high value and varies about that value for relatively long periods. The data thereby suggest that as depth is increased the magnitude of the medium inhomogeneities increase and the size (temporal extent) of the inhomogeneous volumes becomes greater as well.

#### C. TEMPORAL CORRELATION

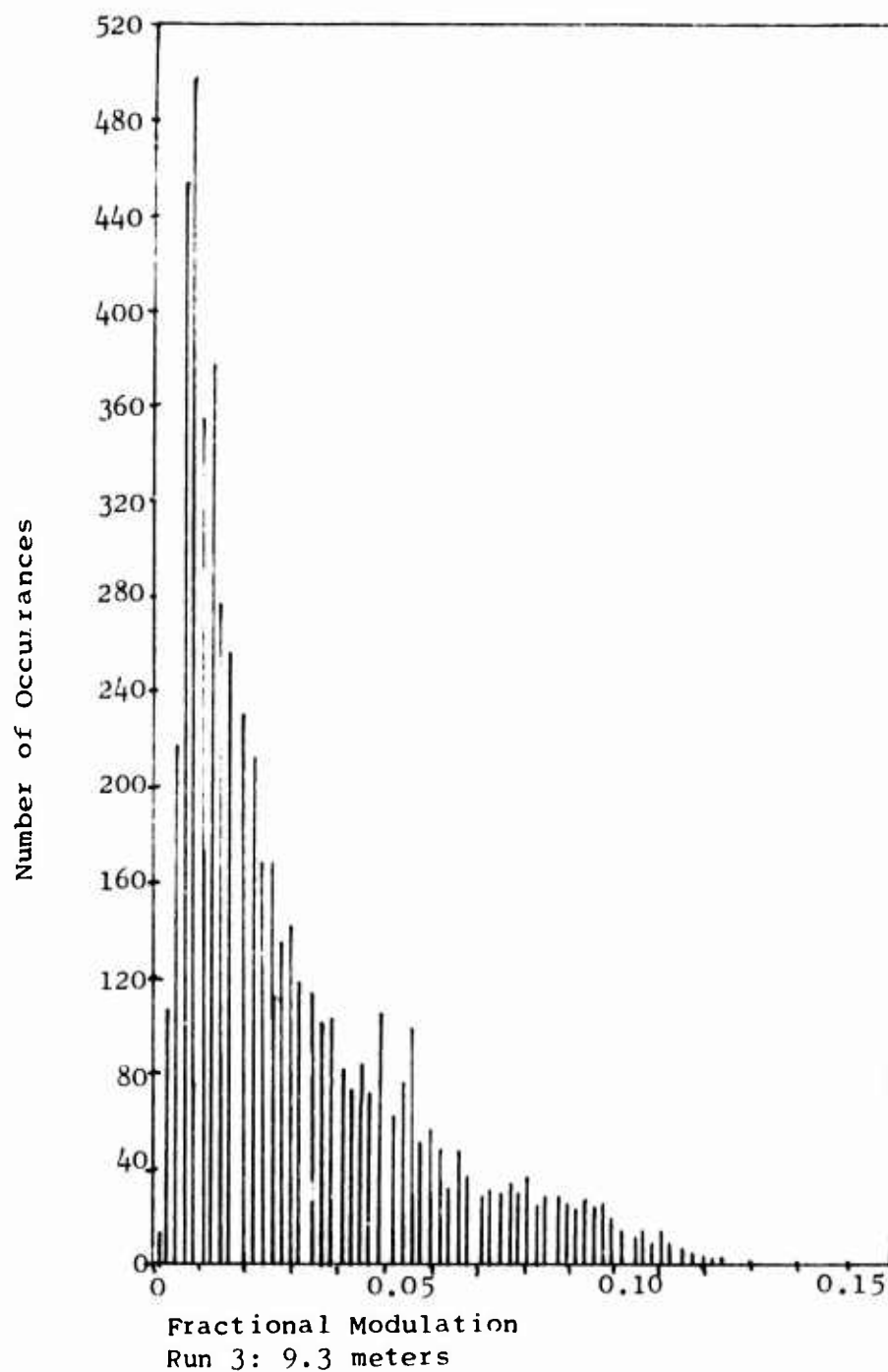
The temporal correlations (time autocorrelations) for each run are shown on Figures V-8 through V-11. The times to 50%





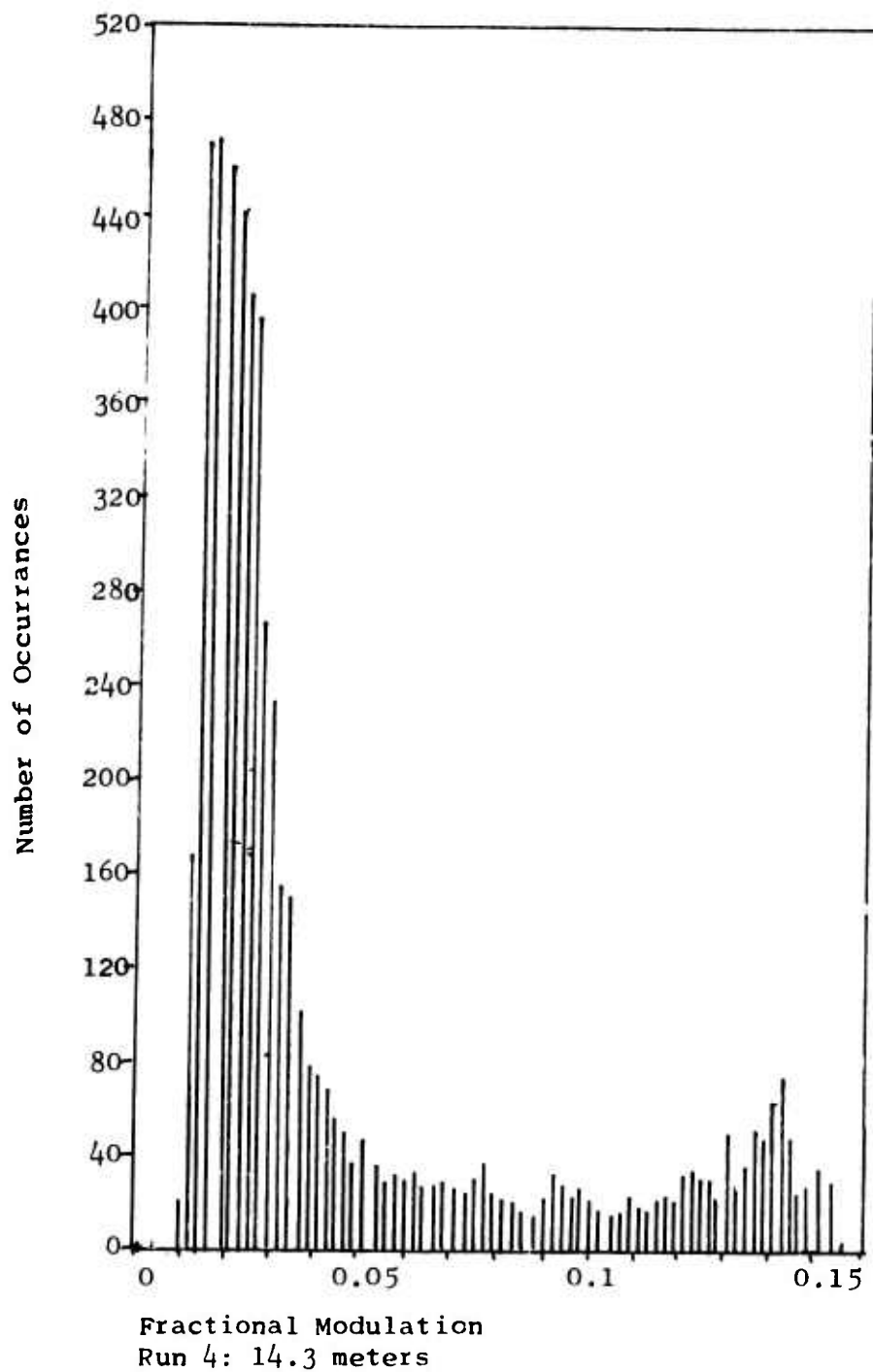
DISTRIBUTION OF FRACTIONAL MODULATION

Figure V-5



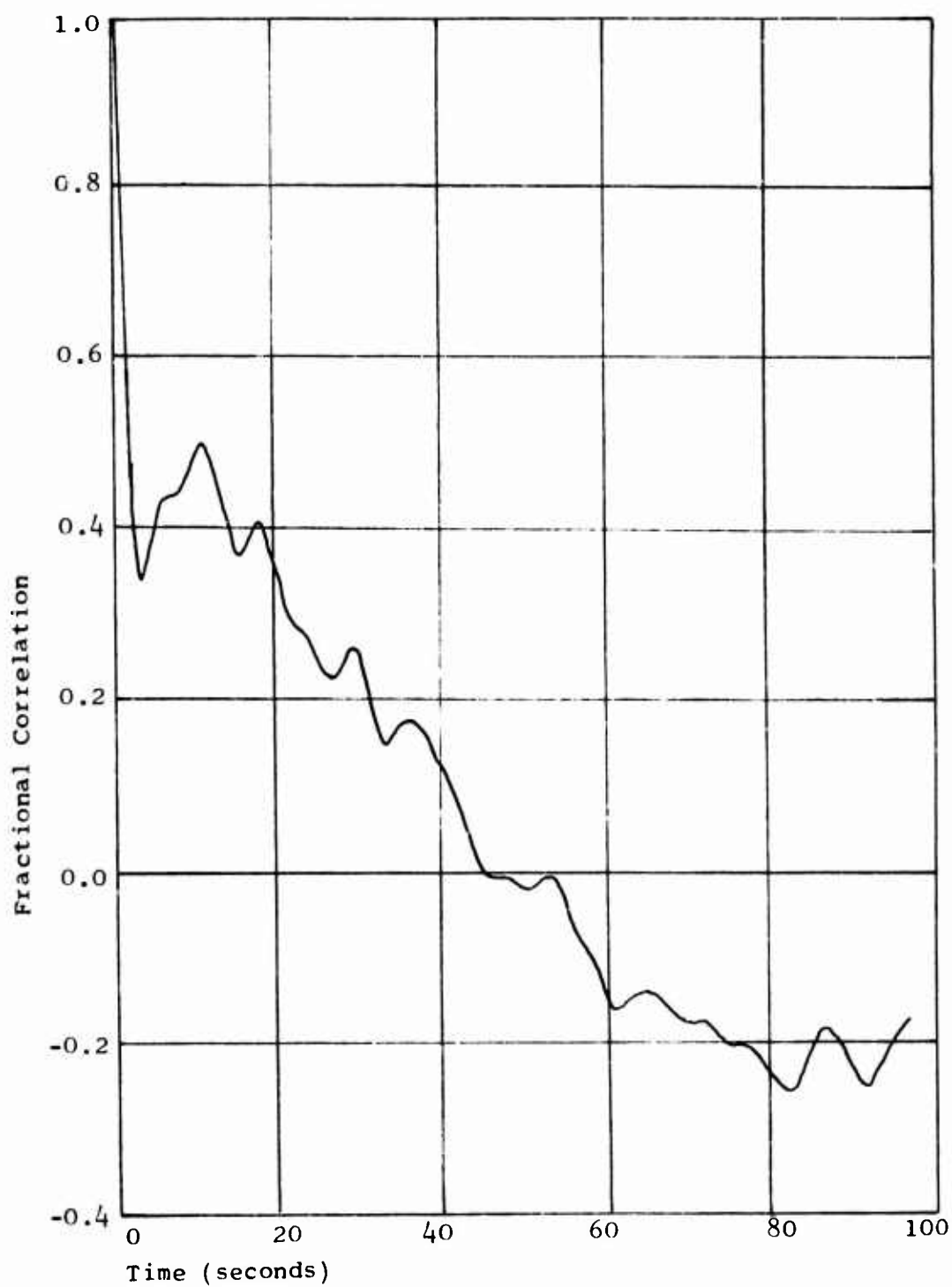
DISTRIBUTION OF FRACTIONAL MODULATION

Figure V-6



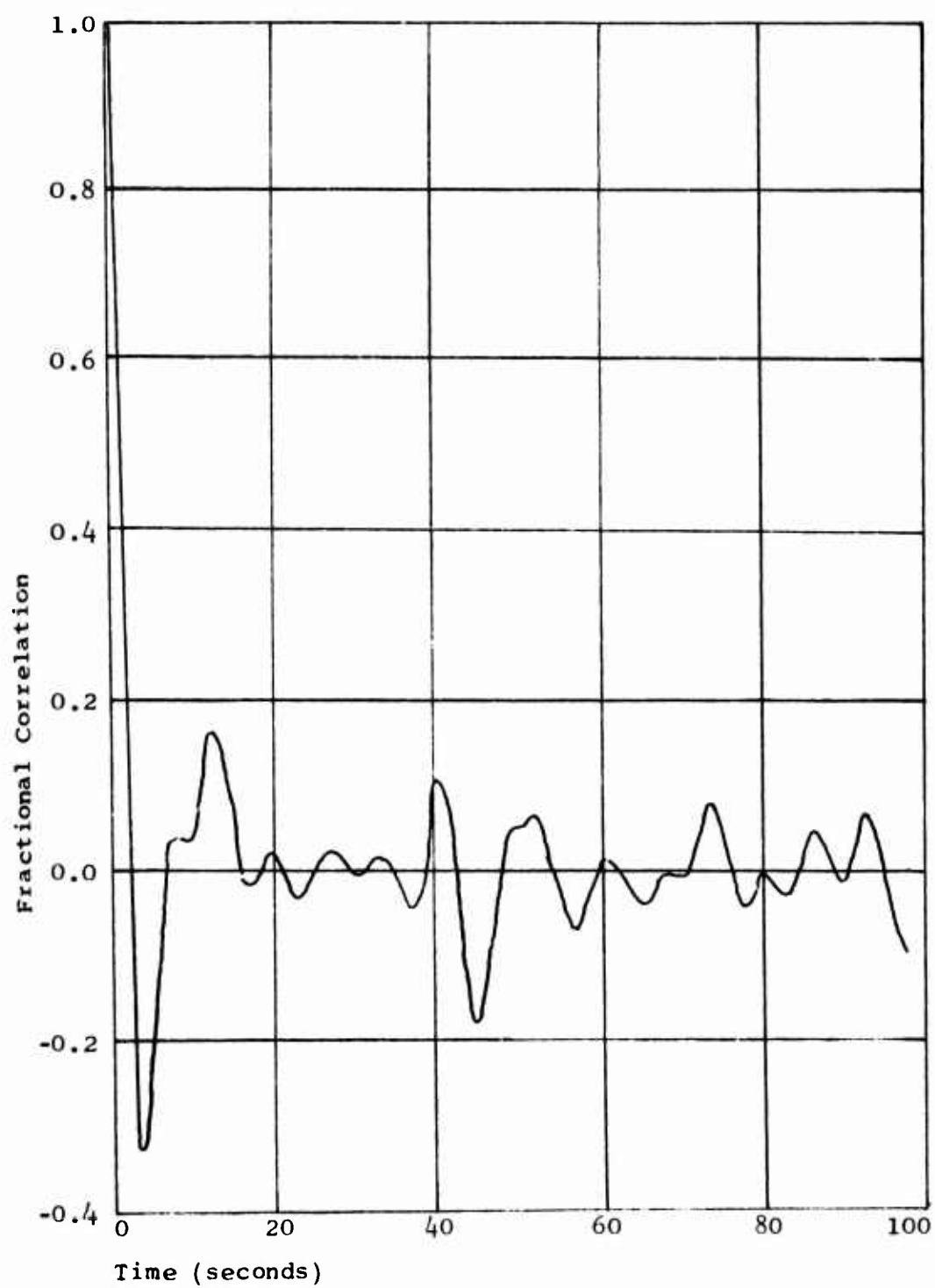
DISTPIBUTION OF FRACTIONAL MODULATION

Figure V-7



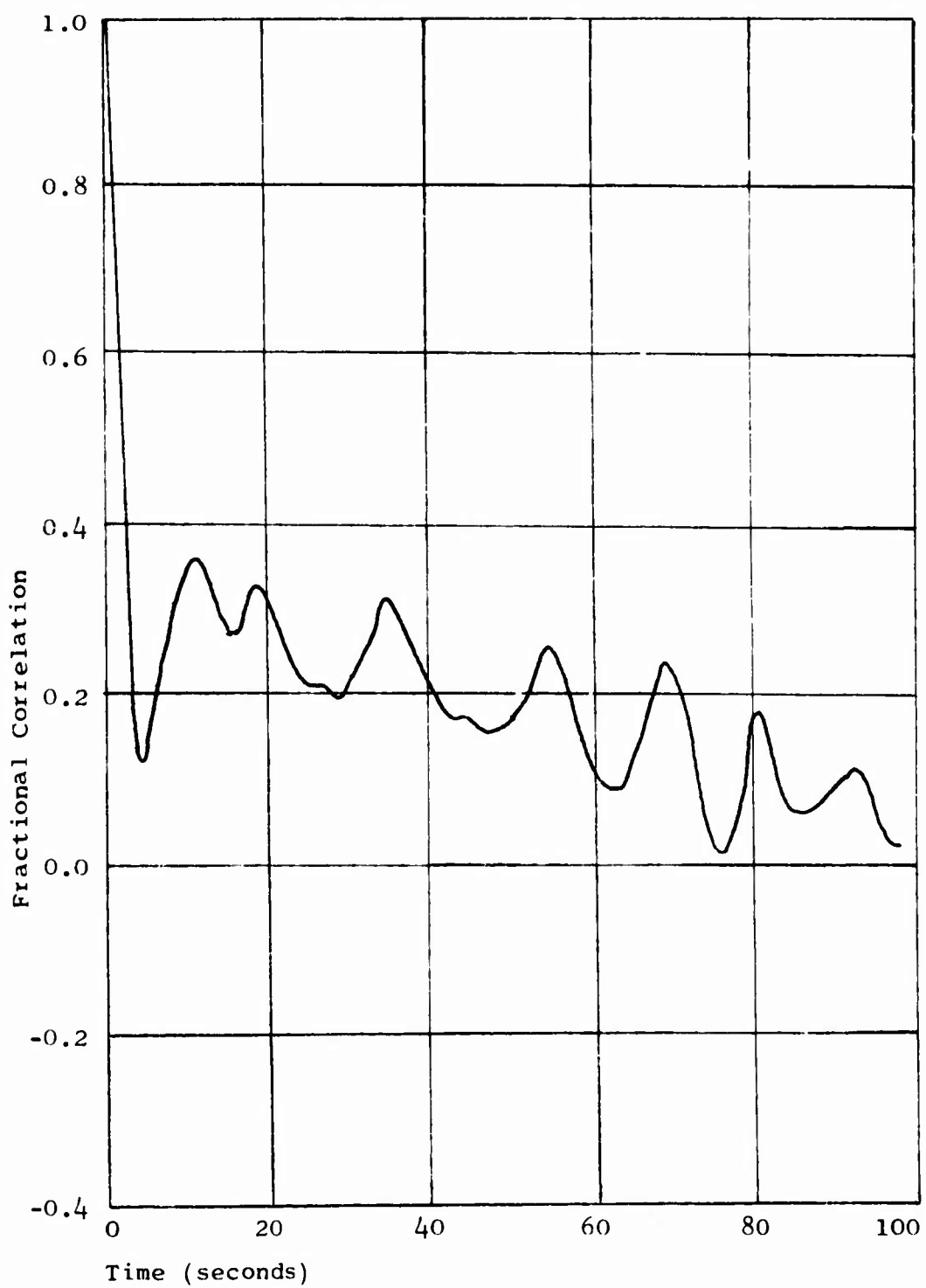
NORMALIZED TEMPORAL CORRELATION OF RUN 2  
DEPTH: 4.3 meters

Figure V-8



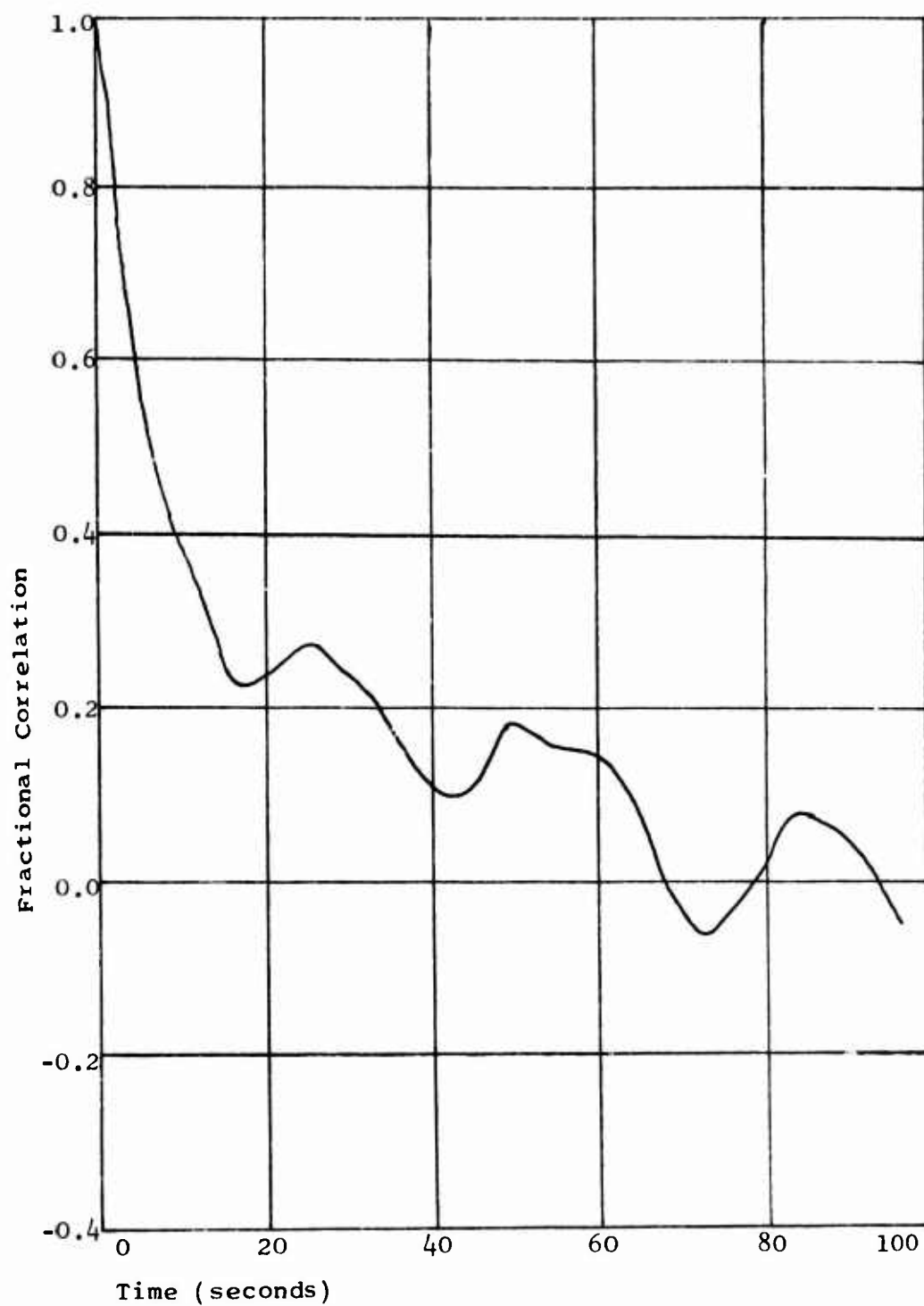
NORMALIZED TEMPORAL CORRELATION OF RUN 5  
DEPTH: 7.3 meters

Figure V-9



NORMALIZED TEMPORAL CORRELATION OF PUN 3  
DEPTH: 9.3 meters

Figure V-10



NORMALIZED TEMPORAL CORRELATION OF RUN 4  
DEPTH: 14.3 meters

Figure V-11

correlation and  $1/e$  correlation are shown below:

Time to 50%:	<u>Depth (meters)</u>	<u>Time (seconds)</u>
	4.3	2.33
	7.3	1.43
	9.3	2.23
	14.3	6.9
Time to $1/e$ :	4.3	3.25
	7.3	1.65
	9.3	2.73
	14.3	11.75

It can be seen from the figures that there is a rapid initial decorrelation down to low as 10% for depths of 4.3 and 9.3 meters, and a slower decorrelation for the 14.3 meter depth. All of these runs show a fluctuation between steadily decreasing minima and maxima with a delayed zero crossing. Whitmarsh, Skudrzyk and Urick (1957) hypothesized that such fluctuation was an indication of the size of the inhomogeneous "patches" around the receiver. If Whitmarsh, Skudrzyk and Urick's hypothesis is correct, then the apparent "patch" sizes show an increase with depth. If this is the case, then small-scale turbulence may tend to dominate at all depths near the surface.

The temporal correlation of the 7.3 meter depth is of interest since it is known that an artificially derived bubble and turbulence field was introduced into the volume under consideration. The 7.3 meter depth shows the most rapid decorrelation to zero of all runs, and twice as frequent fluctuations about zero as for the undisturbed



water at other depths. This suggests that the ambient "patches" of water were chopped into smaller patches by the boat's propellor.

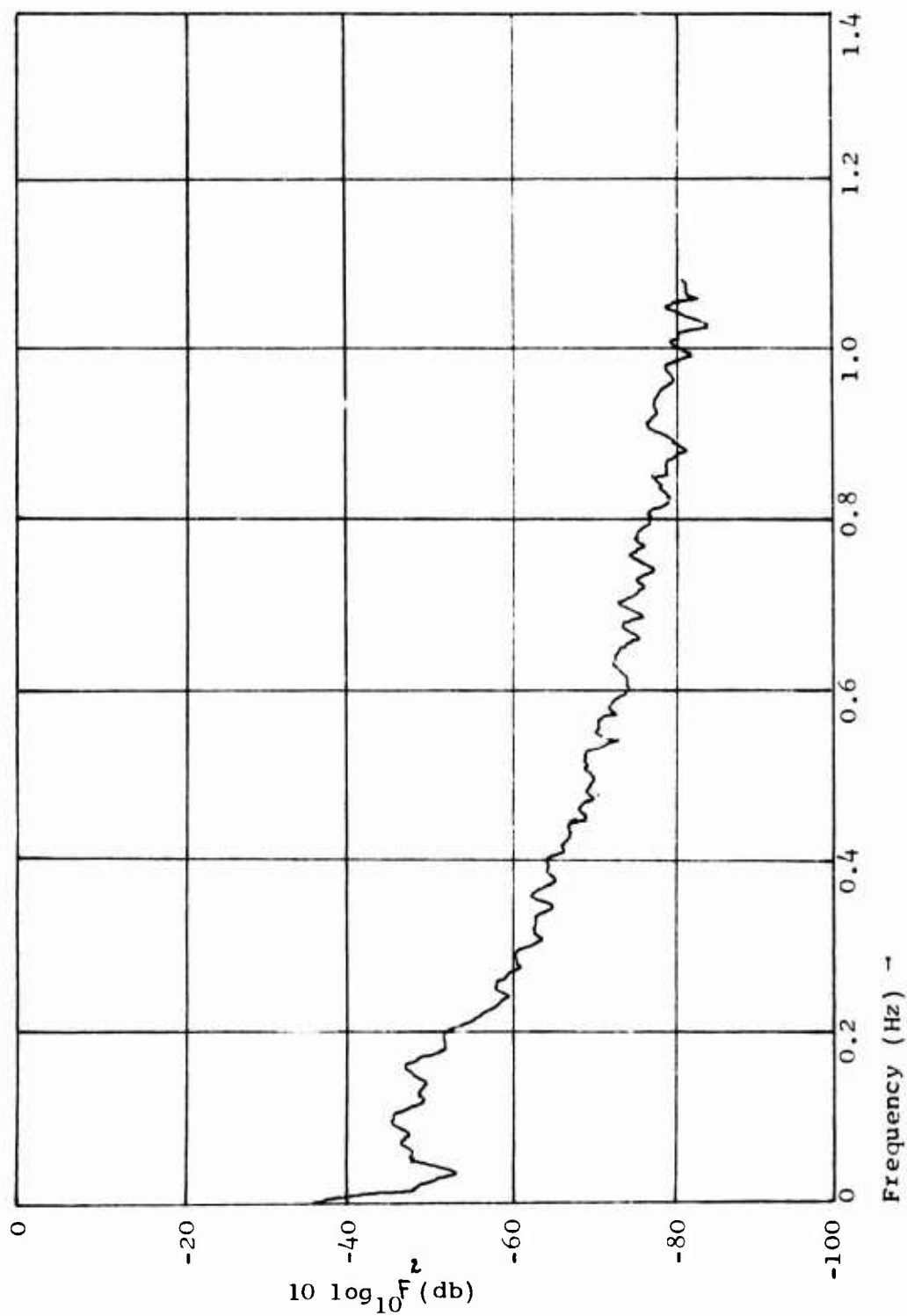
Another hypothesis might be that the inhomogeneous volumes are more or less "frozen" and that the spatial displacement of these volumes about the acoustic axis may cause the appearance of temporal variation. Corroboration of these hypothetical explanations awaits the complete analysis of the data gathered by the oceanographic sensors.

#### D. POWER SPECTRAL DENSITY

The power spectral densities (PSD) of each run are plotted against frequency on Figures V-12 through V-15. The power spectral densities show the highest values at those frequencies (0.08 to 0.12 Hz) which appear to correspond with the observed ocean swell frequencies prevailing during each run.

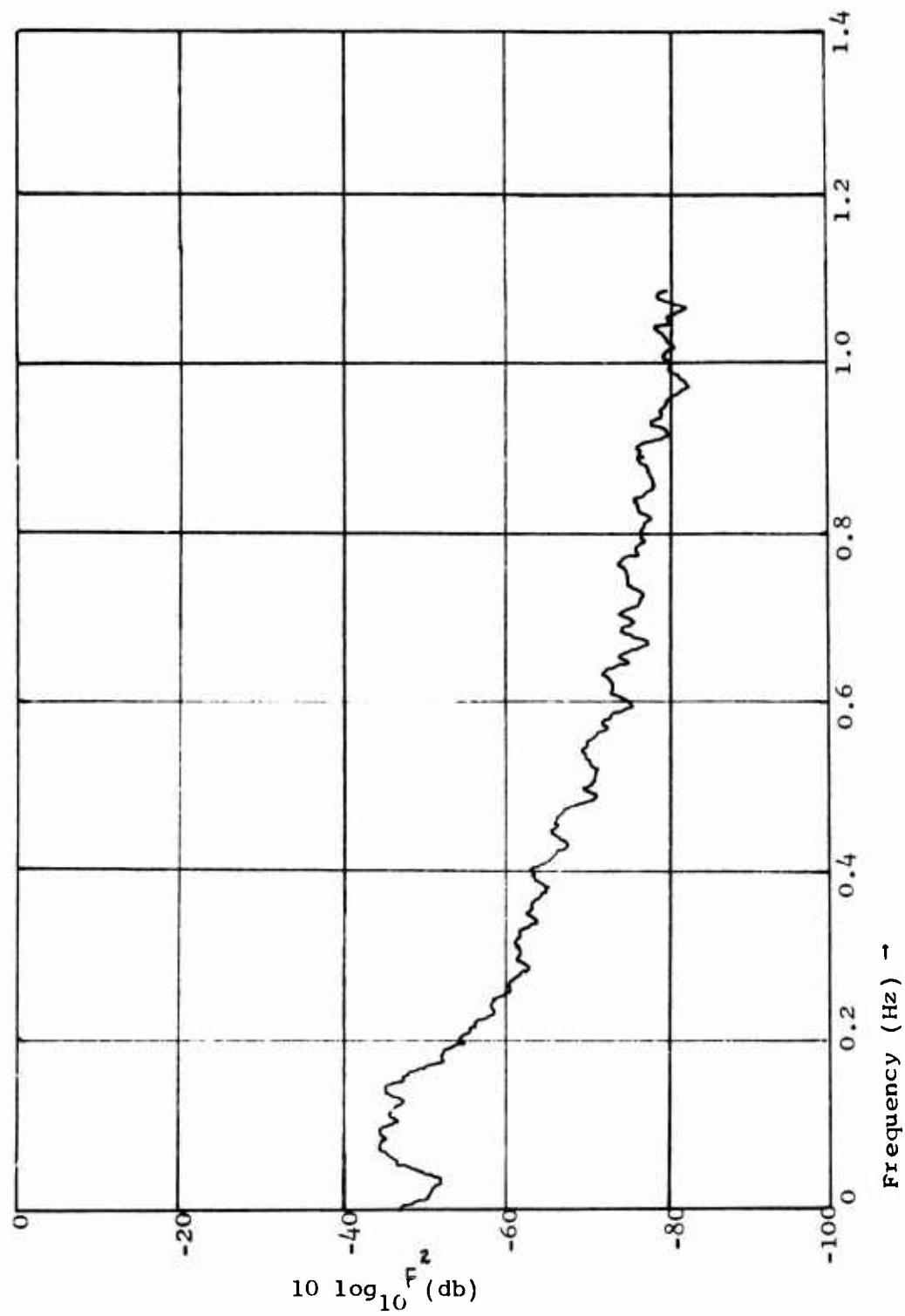
The PSD's show a diminishing value both with increasing frequency and with increasing depth. The two shallowest depths (4.3 and 7.3 meters) show a similarity in the values of PSD from 0 to 1.0 Hz, except for a high "DC" value in the plot for 4.3 meters, resulting in a higher contribution to the integrated value of PSD ( $\sigma^2$ ) at that depth. A similar likeness is apparent between the two greater depths (9.3 and 14.3 meters). For frequencies greater than 0.1 Hz, the PSD at the two shallower depths appear to be consistently four to five db greater than for the deeper depths.

The frequencies of relative maxima along each of the PSD curves show similarities at each end of the spectrum. At the low frequency end, the ocean swell frequencies are more pronounced at



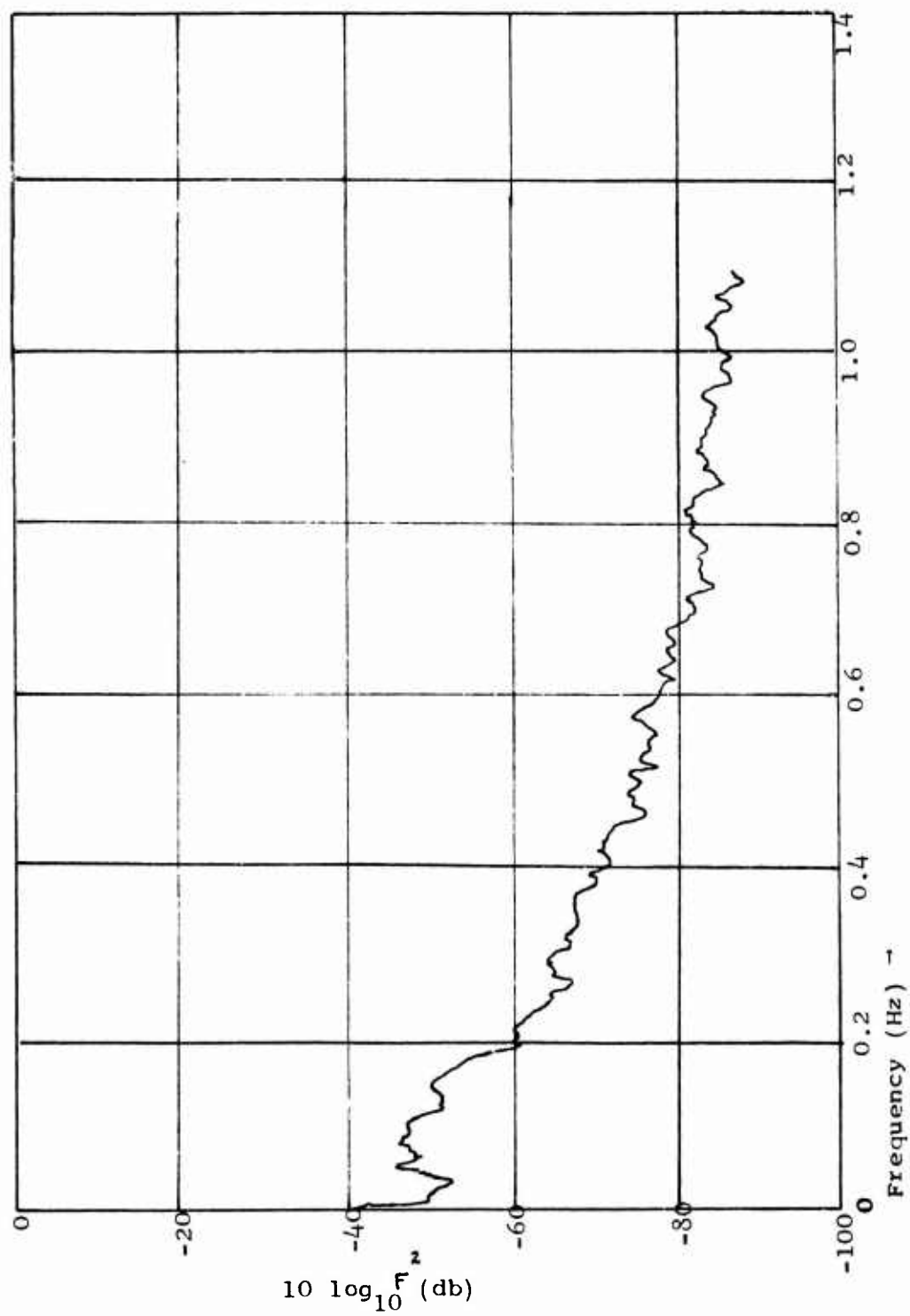
POWER SPECTRAL DENSITY OF RUN 2 DEPTH: 4.3 meters

Figure V-12



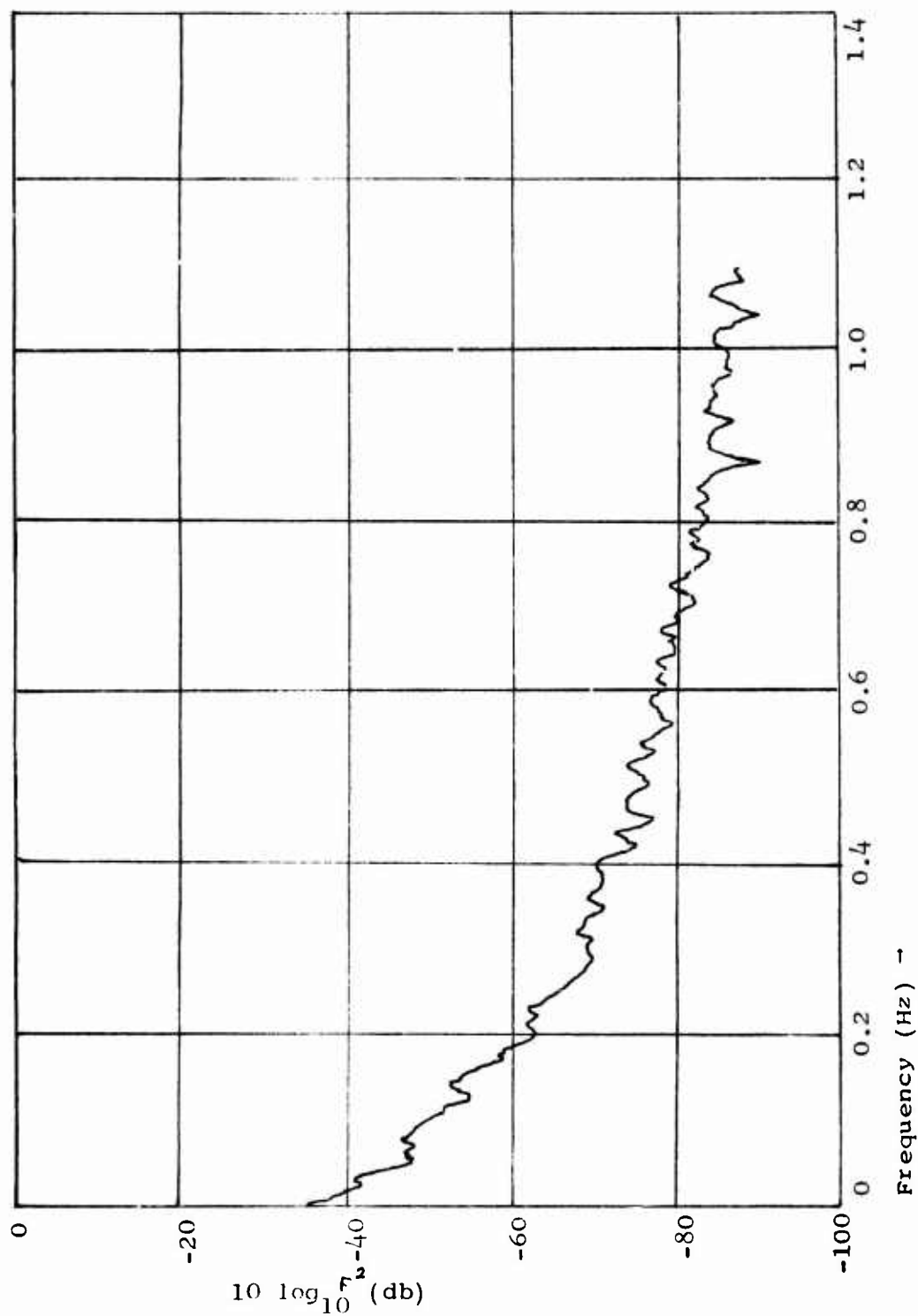
POWER SPECTRAL DENSITY OF RUN 5 DEPTH: 7.3 meters

Figure V-13



POWER SPECTRAL DENSITY OF RUN 3 DEPTH: 9.3 meters

Figure V-14



POWER SPECTRAL DENSITY OF RUN 4 DEPTH: 14.3 meters

Figure V-15

the shallower depths of 4.3 and 7.3 meters. These maxima are higher for the 7.3 meter run than for the 4.3 meter run, which may again show the effect of the boat in the area injecting bubbles and turbulence into the medium. At 9.3 and 14.3 meters, these low frequency maxima tend to dissappear into the overall curve, showing a decrease of the component of the amplitude modulation due primarily to the turbulent effect of surface waves.

The amplitude modulation at the higher frequencies (0.2 to 1.0 Hz) could be due to spatial distributions of temperature, salinity or bubbles which are seen as temporal frequencies due to movement across the acoustic axis.

#### E. CONCLUSIONS

The conclusions presented here await the additional information that the data from the ocean sensors will provide. This information will be available in the theses of LCDR Duchock and LT's Seymour and Bordy, to be presented at the Naval Postgraduate School in March, 1972. The present conclusions are based upon the acoustic data alone and are listed below:

1. Amplitude modulation near the ocean surface ranged in value from zero to sixteen per cent. The maximum amplitudes were not noticably restricted to any one depth, but as depth increased, the distribution of modulation values tended to center about the lower values.

2. The temperal correlation curves for the 4.3, 9.3, and 14.3 meter runs indicate that the size of the volumes of inhomogenieties ("cells") increases as the depth is increased. The correlation

curve for the 7.3 meter run shows a significant decrease in the size of the volumes, a direct result of having a motor launch in the area injecting turbulence and bubbles into the volume of water under investigation. Thus it may be seen that in the ambient ocean, cells of microstructure are larger at greater depth. Near surface cells would tend to be reduced in size by the action of turbulence arising from the effect of surface waves.

3. A comparison of the variances of the two shallow runs, at 4.3 and 7.3 meters, shows that there is a small difference between the two. ( $\sigma^2(4.3) = 0.0013$  and  $\sigma^2(7.3) = 0.0010$ .) Comparing the temporal correlations for the two runs it may be seen that the decorrelation time for the 7.3 meter run is one-half that of the 4.3 meter run, with more numerous fluctuations about the zero axis. Such data suggests a reduction in the size of the cells of inhomogeneity.

Mintzer's formula for the variance of acoustic amplitude modulation, derived for an exponential correlation function (see Appendix I)

$$(CV)^2 \sqrt{\pi} k_0^2 \alpha^2 a r$$

shows the dependence of variance on the wave number,  $k$ , range,  $r$ , the RMS variation of the index of refraction,  $\alpha$ , and the size of the inhomogeneities,  $a$ . In this experiment, wave number and range were constant; since neither  $\sigma^2$  nor  $(CV)^2$  changed appreciably, it can be deduced that there was a simultaneous increase in  $\alpha$  (with the reduction of  $a$ ) for the 7.3 meter run.

4. The parameters which could cause a change in the RMS variation of the index of refraction,  $\alpha$ , were temperature and salinity microstructure, and bubbles.

The effect of resonant bubbles on the change in  $\alpha$  would be very great. Since bubble populations decrease with depth, this might account for the implied decrease in  $\alpha$  with depth. When in the presence of a motor launch, the size of the inhomogeneous volumes were made smaller, there was a concurrent increase in the bubble population and in the value of  $\alpha$ . The changes and variations in the other parameters, chiefly temperature and salinity, should be accounted for in subsequent analyses, and a quantitative estimate of the relative effect of bubbles on amplitude modulation may be possible.

5. The power spectral densities indicate that most of the modulation takes place at those frequencies corresponding to the frequencies of the surface waves. This effect was noticable down to the greatest depth (14.3 meters) although the prominence of the PSD's at the surface-wave frequencies tended to disappear with greater depth.

6. In summary, the major causes of acoustic amplitude modulation at 60 kHz in this near-surface volume of ocean could well have been the effects of mobile, random bubble populations acted upon by particle velocities generated by the action of surface waves.

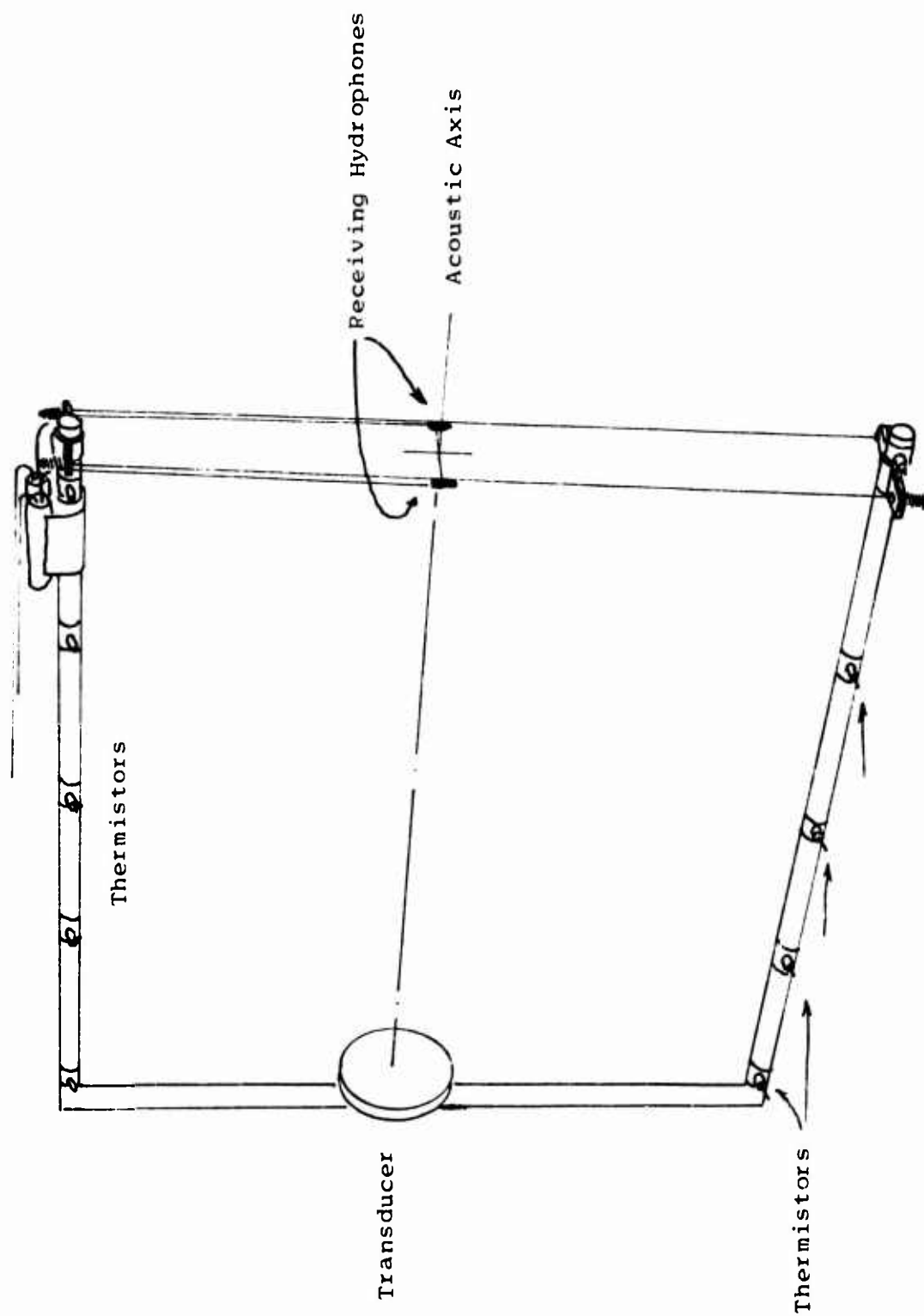
#### F. SUGGESTIONS FOR FURTHER EXPERIMENTS

The basic questions left incompletely answered by this investigation are the relative effect of bubbles versus the other causes of amplitude modulation near the surface, and the spatial effect of the inhomogenieties upon the transmission of sound. Later analysis of the oceanographic data will help to answer this question.



It is suggested that further experiments of this kind would be valuable. In order to investigate the spatial effects of the inhomogenieties, the basic apparatus of this experiment could be modified as shown in Figure V-16 in order to attempt some cross-correlation of the amplitude modulation received by each of the hydrophones. In order to more completely investigate the question of transport versus the stationary oscillation of inhomogenieties, more thermistors could be mounted as shown on Figure V-16.

The question of bubbles and bubble populations in an area might be answered either by analysis of all other parameters and their effect, or by deliberately generating bubbles in the volume of water to note their effect on ambient conditions.



SUGGESTED CONFIGURATION OF THE ACOUSTIC APPARATUS  
FOR SPATIAL STUDY

Figure V-16

## APPENDIX

### THE RELATION BETWEEN F AND CV

The relation between F, the fractional modulation which is used in the development of the experiment, and CV, which Stone and Mintzer call the "coefficient of variation" will be developed.

Fractional modulation is defined in terms of the voltages as shown below:

$$F = \left[ \frac{V_{MAX} - V}{V_{MAX}} \right]$$

where V = modulated voltage

V<sub>MAX</sub> = unmodulated voltage

The coefficient of modulation is defined by Stone and Mintzer in terms of pressure amplitude, P, as shown below:

$$(CV)^2 = \frac{(\langle P^2 \rangle - \langle P \rangle^2)}{\langle P \rangle^2}$$

Then if one assumes that  $P = KV$ , or that a constant factor relates the two values, as it does in the equipment used for these experiments, then:

$$(CV)^2 = \frac{(\langle (KV)^2 \rangle - \langle KV \rangle^2)}{\langle KV \rangle^2} = \frac{K^2}{K^2} \left[ \frac{\langle V^2 \rangle - \langle V \rangle^2}{\langle V \rangle^2} \right]$$

Fractional modulation was used in this experiment because small fluctuations were anticipated. A development between the variance of the fractional modulation and the term  $(CV)^2$  is shown below:

$$F = \left[ \frac{V_{MAX} - V}{V_{MAX}} \right] = \left( 1 - \frac{V}{V_{MAX}} \right)$$

$$\sigma^2 = [\langle F^2 \rangle - \langle F \rangle^2]$$

$$\sigma^2 = \langle (1 - \frac{V}{V_{MAX}})^2 \rangle - \langle 1 - \frac{V}{V_{MAX}} \rangle^2$$

$$\sigma^2 = \langle 1 - 2\frac{V}{V_{MAX}} + (\frac{V}{V_{MAX}})^2 \rangle - (\langle 1 \rangle - \langle \frac{V}{V_{MAX}} \rangle)^2$$

$$\sigma^2 = \langle 1 \rangle - 2\langle \frac{V}{V_{MAX}} \rangle + \langle (\frac{V}{V_{MAX}})^2 \rangle - \langle 1 \rangle + 2\langle \frac{V}{V_{MAX}} \rangle - \langle \frac{V}{V_{MAX}} \rangle^2$$

$$\sigma^2 = \langle (\frac{V}{V_{MAX}})^2 \rangle - \langle \frac{V}{V_{MAX}} \rangle^2$$

or, since  $V_{MAX}$  is a constant,

$$\sigma^2 = \frac{1}{(V_{MAX})^2} [\langle V^2 \rangle - \langle V \rangle^2]$$

multiplied by the factor  $\frac{\langle V \rangle^2}{\langle V \rangle^2}$  the variance may now be expressed:

$$\sigma^2 = \frac{\langle V \rangle^2}{(V_{MAX})^2} \left[ \frac{\langle V^2 \rangle - \langle V \rangle^2}{\langle V \rangle^2} \right]$$

Where the second term is a numerical value multiplied by Mintzer and Stones' expression for the coefficient of variation.

The value multiplied against the coefficient of variation is

$$\frac{\langle V \rangle^2}{(V_{MAX})^2}$$

which is numerically obtainable as follows:

$$\langle F \rangle = 1 - \langle V \rangle$$

$$\therefore \langle V \rangle = 1 - \langle F \rangle$$

for all runs, the value of  $V_{MAX}$  was 6 volts.

$$\therefore \frac{\langle V \rangle^2}{(V_{MAX})^2} = \frac{(1 - \langle F \rangle)^2}{36}$$

The relation between  $\sigma^2$  and  $(CV)^2$  will be calculated for the 4.3 meter run:

$$\frac{\langle v \rangle^2}{(V_{MAX})^2} = \frac{(1 - 0.0526)^2}{36} = \frac{(0.9474)^2}{36} = 0.02496$$

and in the 7.3 meter run was:

$$\frac{\langle v \rangle^2}{(V_{MAX})^2} = \frac{(1 - 0.0460)^2}{36} = \frac{(0.9540)^2}{36} = 0.02530$$

It is noted that the conversion factor does not significantly change with depth, and therefore conclusions about the relative value of  $(CV)^2$  at two different depths are applicable as well as the relative value of  $\sigma^2$  at these depths.

```

C      PROGRAM 1
C      PROGRAM CONVERT(INPUT,OUTPUT,PUNCH)
C      AVGX IS THE AVERAGE DISTANCE THE RECORDING
C      DRUM ADVANCES PER ONE HOUR REAL TIME, I.F. THE DATA
C      SAMPLING RATE,
      DIMENSION U(6000),V(6000),N(80),IBUFF(5000),NK(80)
      IK=0
      READ 97,L
97     FORMAT(I2)
      READ 98,AVGX
98     FORMAT(F10.0)
      READ 96,IFILE
96     FORMAT(I2)
      DO 116 JJ=1,L
      DELT=36.0/AVGX
      PRINT 99,DELT
99     FORMAT(1H0,2X,25H SAMPLING INTERVAL EQUALS, 1X,E15.7,
11X,7HSECONDS)
      PRINT 700,JJ
700    FORMAT(1H0,9HJJ EQUALS, 2X,I2)
      DO 100 I=1,5500
      U(I)=0.0
100    V(I)=0.0
101    DO 202 I=1,5000
202    IBUFF(I)=0
      COUNTX=0.0
      COUNTY=0.0
      NUM=5000
      M=1
      K8=0
      CALL LIOF(5LRBCD1,IBUFF,NUM,NPAR,NEOF)
      IF(NEOF) 602,200
200    K=-7
      KF=0
201    K=K+8
      KA=K+8
      KC=0
      DO 102 KB=K,KA
      KC=KC+1
      NK(KC)=IBUFF(KB)
102    IF (NK(KC).EQ.0) KF=1
      DECODE(80,103,NK) (N(I),I=1,80)
103    FORMAT(80R1)
      DO 104 I9=1,80
      IF(N(I9).EQ.50B) GO TO 106
      IF(N(I9).EQ.55B) GO TO 107
      IF(N(I9).EQ.34B) GO TO 108
C      SYMCOL / (50) REPRESENTS AN INCREMENT TRAVEL IN THE
C      MINUS X OR Y DIRECTION BY THE DIGITIZER.
C      SYMBOL 0,(55B), REPRESENTS A ZERO INCREMENT TRAVEL
C      IN THE X OR Y DIRECTION BY THE DIGITIZER
C      SYMBOL 1,(34B), REPRESENTS AND INCREMENT TRAVEL IN
C      THE POSITIVE X OR Y DIRECTION BY THE DIGITIZER
C      SYMBOL *,(47B), IS A FLAG INSERTED IN IN THE RECORD
C      BY THE PERSON DIGITIZING BY USE OF THE IRG
      GO TO 104
106    RX=-0.01
      K8=K8+1
      GO TO 109
107    RX=0.0
      K8=K8+1
      GO TO 109
108    RX=0.01
      K8=K8+1
109    K3=K8/2
      K3=2*K3
      IF(K3.EQ.K8) GO TO 111
      COUNTX=COUNTX+RX
      IF(COUNTX.NE.0.0) GO TO 110
      GO TO 104
110    U(M)=COUNTX
      GO TO 104

```

```

111 COUNTY=COUNTY+RX
    IF(COUNTX.NE.0.0) GO TO 112
    GO TO 104
112 V(M)=COUNTY
    COUNTX=0.0
    M=M+1
    IF(M.GT.6000) GO TO 600
104 CONTINUE
    IF(KF.EQ.1) GO TO 113
    GO TO 201
113 MAX=M-7
105 PRINT 205,M,19
205 FORMAT(1H0,2I10)
C   TOTAL TIME OF THE RECORD EQUALS NUMBER OF DATA
C   POINTS TIMES THE SAMPLING INTERVAL, DELT.
    TIME=(M*DELT)/3600.0
    PRINT 115,TIME
115 FORMAT(1H,10X,26H TOTAL TIME OF RECORD EQUALS,2X,
1E15.7,2X,7H HOURS.)
    PRINT 215,(V(I),I=1,M)
    PRINT 215,(U(I),I=1,M)
215 FORMAT(1H,10X,14F7.2)
    PUNCH 213,(V(I),I=1,M)
213 FORMAT(14F5.2)
    PUNCH 2100
2100 FORMAT (20H***** )
116 CONTINUE
    STOP
600 PRINT 601
601 FORMAT(1H0,20X,37H***** U AND V SPACE INADEQUATE ***,)
602 IK=IK+1
    IF(IK.GT.IFILE)603,200
603 STOP
    END

```





```

801 FORMAT(6F10.5)
803 FORMAT(40H LONG WAVE STUDY OF MONTEREY BAY ,5X,7H DATE ,2A
14,10H HOUR ,2A4//,5X,10A4//)
C BAND WIDTH FREQUENCIES OF TOTAL ENERGY FLUX- CMIN,CMAX
CMIN = 0.0
CMAX = 0.2

C COMPUTING POWER SPECTRUM
PI=3.14159265
FB = FBHZ*2.0*PI
FE = FEHZ*2.0*PI
DF = DH7 *2.0*PI
FMIN = .0*PI*CMIN
FMAX = .0*PI*CMAX
900 FORMAT(14F5.2)
READ(5,900)(F1(I), I=1,NTS)
XMAX = F1(1)
DO 23 I=2,NTS
IF(XMAX.GE.F1(I)) GO TO 23
XMAX = F1(I)
23 CONTINUE
DO 24 I=1,NTS
F1(I) = ((F1(I)-XMAX)*(-.212))
24 CONTINUE
WRITE(6,901)(F1(I), I=1,NTS)
901 FORMAT(14F7.2)
CALL TREND(F1,NTS,DT,CALX1)
DO 10 M=1,MLAG
SUM=0.0
NMAX=NTS-M+1
DO 8 I=1,NMAX
NN=M+I-1
8 SUM=SUM+F1(I)*F1(NN)
XX=M-1
XNMAX=NMAX
TAU(M)=XX*DT
PHI(M)=SUM/XNMAX
WRITE(6,300)(PHI(1))
300 FORMAT(5X,'PHI(1)=' ,F20.5)
PHN(M) = PHI(M)/PHI(1)
10 CONTINUE
CALL PARZ(MLAG,PHI)
798 FORMAT('PHI(0)=' ,F10.4//)
NFREQ=(FE-FB)/DF+0.1
DO 14 N=1,NFREQ
XN=N
FREQ(N)=(XN-1.0)*DF+FB
14 CYCL(N) = FREQ(N)/(2.0*PI)

```



```

109 FORMAT (I6)
888 RETURN
END

C SUBROUTINE PRESS(FREQ,SPEC,NFREQ,H,X2,DF,FB,FMAX)
DIMENSION FREQ(NFREQ),SPEC(NFREQ)
NMAX = (FMAX-FB)/DF+1.0
DO 11 I=1,NMAX
C CALCULATE LINEAR WAVE LENGTH BY NEWTONS METHOD
IF(FREQ(I)-0.00001) 8,9,7
7 XKHO = FREQ(I)*FREQ(I)*H/9.80
IF(XKHO-6.3) 5,1,1
1 XKH = XKHO
5 GO TO 9
3 XKH = SQRT(XKHO)
SH = SINH(XKH)
CH = COSH(XKH)
EPS = XKHO-XKH*SH/CH
SLOPE = -EPS/SLOPE
DXKH = -EPS/SLOPE
IF(ABS(DXKH/XKH)-0.0001) 9,9,4
4 XKH = XKH+DXKH
8 RESPF = 1.00
9 XKH = XKH/H
RESPF = COSH(XK*H)/(COSH(XK*(H-X2)))
IF(RESPF-10.0) 11,11,12
11 SPEC(I) = RESPF*RESPF*SPEC(I)
12 RETURN
END

SUBROUTINE TREND(FX,NTS,DT,CALXX)
DIMENSION FX(NTS)
C CALIBRATION RECORD
DO 104 I=1,NTS
104 FX(I) = FX(I)*CALXX
C COMPUTING THE LINEAR TREND
FNTS = NTS
SUMF = 0.0
DO 101 I=1,NTS
101 SUMF = SUMF + FX(I)
SUMF1 = 0.0
DO 102 I=1,NTS
102 XI = I

```

```

102 SUMF1 = SUMF1 + XI*FX(I)
    XNMI = NTS-1
    XNPI = NTS+1
    XM = (1.0/DT)*(12.0*SUMF1/(FNTS*XNMI*XNPI)-6.0*SUMF/(XNMI*FNTS))
    B = SUMF/FNTS-XM*XNPI*DT/2.0
    FMEAN = SUMF/FNTS
    WRITE(6,9) FMEAN,XM,B
    9 FORMAT (3X,8H MEAN = ,F10.5,3X,9H SLOPE = ,F10.5,3X,13H INTERCEPT
    1 = ,F10.5//)
    DO 103 I=1,NTS
    103 FX(I) = FX(I) - (B+XM*XI*DT)
    RETURN
END

```

```

SUBROUTINE SMO(MD,X1,X2,NFREQ)
DIMENSION X1(MD),X2(MD)
DO 1 N=1,MD
    NA=N+MD
    NN=NFREQ-N+1
    NB=NN-MD
    X2(N) = 0.25*(X1(1)+X1(NA))+0.5*X1(N)
    X2(NN) = 0.5*(X1(NN)+X1(NB))
    1 MB=MD+1
    ME=NN-1
    DO 2 N=MB,ME
    2 NA=N+MD
    NB=NN-MD
    X2(N) = 0.25*(X1(NA)+X1(NB))+0.5*X1(N)
    RETURN
END

```

C

```

SUBROUTINE PARZ(MLAG,PHI)
PARZ SUBROUTINE PARZEN FILTERS AUTO-CORRELATION FUNCTION
DIMENSION PHI(MLAG)
XMLAG = MLAG
MLAGH = XMLAG/2.0-0.1
MLAGH1 = MLAGH + 1
DO 31 M=1,MLAGH
    MM = M-1
    RM = MM
    UM = R/XMLAG
    UM = 1.0-6.0*RM*RM*(1.0-RM)
    PHI(M) = PHI(M)*UM
31 CONTINUE
DO 32 M = MLAGH1,MLAG

```

```
MM = M-1  
R = MM  
RM = R/XMLAG  
RM1 = (1.0-RM)  
UM = 2.0*RM1*RM1*RM1  
PHI(M) = PHI(M)*UM  
32 CONTINUE  
RETURN  
END
```

```

C      PROGRAM IV
C      PROGRAM TO CALCULATE THE VARIANCE OF THE FRACTIONAL
C      MODULATION OF EACH RUN.
      DIMENSION X(5600)
      READ (5,1100)(X(I),I=1,5600)
1100  FORMAT(14F5.2)
      XMAX=X(1)
      DO 23 I=2,5600
      IF (XMAX.GE.X(I)) GO TO 23
      XMAX=X(I)
23  CONTINUE
      DO 24 I=1,5600
      X(I)=(X(I)-XMAX)*(-.212)
24  CONTINUE
      X3=0.0
      DO 30 I=1,5600
      X3=X3+(X(I)**2)
30  CONTINUE
      X3=X3/5600
      WRITE(6,99)(X3)
99  FORMAT(5X,'MEAN SQUARE=',F5.4)
      X4=0.0
      DO 31 I=1,5600
      X4=X4+X(I)
31  CONTINUE
      X4=X4/5600
      WRITE(6,100)(X4)
100  FORMAT(5X,'MEAN=',F5.4)
      X5=X4**2
      X6=X3-X5
      WRITE(6,101)(X6)
101  FORMAT(5X,'SIGMA SQUARE=',F5.4)
      STOP
      END

```

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